# **Predictive Control** Lecture 2

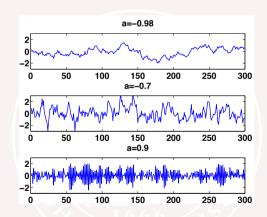
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# Some Stochastic Signals with unit variance



The signals seem to have different properties We should design different predictors/controllers for them How do we describe spectral properties of stochastic signals? Lecture 2

#### Today's lecture

- Stochastic Signal Models (Chapter 3)
- Recursive Identification (Chapter 4)

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# **Stochastic Variables**

- $\mu_x = E(x)$  = mean value of x
- $\sigma_x^2 = V(x) = E(x \mu_x)^2$  = variance of x
- $Cov(x, y) = E(x \mu_x)(y \mu_y)$  the cross-covariance between x and y
- Uncorrelated if Cov(x, y) = 0

If x and y are (column) vectors

- $P_x = V(x) = E(x \mu_x)(x \mu_x)^T$  = variance matrix
- $Cov(x, y) = E(x \mu_x)(y \mu_y)^T$  cross-covariance matrix

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# White noise

A sequence of uncorrelated stochastic variables with

$$E(w_i) = 0,$$
  $E(w_i^2) = \sigma^2,$   $Cov(w_i, w_j) = 0, i \neq j$ 

is called **white noise** with variance  $\sigma^2$ .

This is in some sense a "worst case" enemy, since old noise observations reveal no information at all about the future.

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# **Autocovariance functions**

For a stationary signal  $y_k$  the covariance function is defined as

$$C_{yy}(\tau) = Cov(y(k+\tau)y(k)^T)$$

2 min-problem: Assume that

$$x_{k+1} = Ax_k + e_k$$

(*e* is white noise) and that x is stationary. Why is  $C_{xx}(\tau) = A^{\tau}V(x)$  for  $\tau \geq 0$ ? What happens if  $\tau < 0$ ? .

# **Useful facts**

If y = Ax + b where x is stochastic and A and b known, then

- $\bullet$  E(y) = AE(x) + b
- $V(x) = AV(x)A^T$

Example: The system  $x_{k+1} = Ax_k + e_k$  with e white noise with variance matrix R has

$$P_x(k+1) = Cov(x_{k+1}x_{k+1}^T) = \dots = AP_x(k)A^T + R$$

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# **Spectrum of stochastic signals**

The **auto-spectrum** of y is defined as  $S_{yy}(z) = \sum\limits_{k=-\infty}^{\infty} C_{yy}(k) z^{-k}$ 

The **power spectral density** at frequency  $\omega = 2\pi f$  is given by

$$S_{yy}(e^{i\omega h})$$

Similar definitions for  $S_{uu}(z)$  and  $S_{yu}(z)$ 

If Y(z) = G(z)U(z) + V(z) with uncorrelated u and v then

$$S_{yy}(z) = G(z)S_{uu}(z)G^{T}(z^{-1}) + S_{vv}(z)$$

# **Example**

Assume |a| < 1, h = 1, and w is white noise (W(z) = 1)

$$y_k = -ay_{k-1} + w_k \iff Y(z) = \frac{1}{1 + az^{-1}}W(z) = \frac{z}{z + a}W(z)$$

$$egin{array}{lcl} C_{yy}(z) & = & rac{1}{1-a^2}(-a)^{| au|} \ S_{yy}(z) & = & rac{1}{1+az}rac{1}{1+az^{-1}} \ S_{yy}(e^{i\omega}) & = & rac{1}{1+a^2+2a\cos\omega} \end{array}$$

Matlab: xcorr and pwelch,

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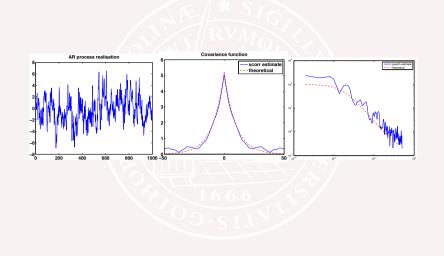
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# **Example**

```
figure(1);
nr=1000; % nr of samples
a=-0.9; h=1; w=randn(nr,1);
y = filter(1, [1 a], w);
plot(y,'linewidth',1.4)
figure(2);
[C,tau] = xcorr(y,50,'unbiased');
plot(tau, C, 'linewidth', 1.4); hold on;
plot(tau,1/(1-a^2)*(-a).^abs(tau),'r--','linewidth',1.4);
figure(3); nrpoints=512;
[Pxx,f]=pwelch(y,[],[],nrpoints,1/h);
loglog(f,Pxx,'linewidth',1.4); hold on;
loglog(f,1./(1+a^2+2*a*cos(2*pi*f)),'r','linewidth',1.4)
```

# **Matlab simulation**



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# **Stationary State Covariance**

Assume A stable and v and e uncorrelated

$$\begin{array}{rcl} x_{k+1} & = & Ax_k + v_k \\ y_k & = & Cx_k + e_k \end{array}$$

Stationary covariance  $P_x = E(xx^T)$  can be found from

$$P_x = AP_xA^T + Q_v$$

Stationary covariance  $P_{\nu}$  is then given by

$$P_y = CP_xC^T + Q_e$$

Matlab: Px = 1yap(A,Qv); Py=C\*Px\*C+Qe

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#### **Real-time Parameter Estimation**

- 1. Introduction
- 2. Least squares and regression
- 3. Dynamical systems
- 4. Experimental conditions
- 5. Examples
- 6. Conclusions

# **Identification Techniques**

- Nonparametric methods
  - Frequency response analysis
  - Transient response analysis
  - Correlation analysis
  - Spectrum analysis
- · Parametric methods
  - Least Squares (LS)
  - Maximum Likelihood (ML)
  - Subspace-model identification
- · Identification for control
- Model approximation/ Model Reduction

Related areas: Statistics, Numerical analysis, Econometrics,

# **System Identification**

- How to get models?
  - Physics (white-box)
  - Experiments (black-box)
  - Combination (grey-box)
- Experiment planning
- · Choice of model structure
  - Transfer functions
  - Impulse response
  - State-space models
- · Parameter estimation
  - System identification/Statistics/Inverse problems
- Validation

## **Least Squares and Regression**

- Introduction
- The LS problem
- Interpretation
  - Geometric
  - Statistic
- Recursive Calculations
- · Continuous time models

. . .

# Good Methods are adopted by Everybody

- Mathematics
- Statistics
- Numerical analysis
- Physics
- Economics
- Biology
- Medicine
- Control
- · Signal processing

**The Least Squares Method** 

The problem: The Orbit of Ceres

The problem solver: Karl Friedrich Gauss

The principle: "Therefore, that will be the most probable system of values of the unknown quantities, in which the sum of the squares of the differences between the observed and computed values, multiplied by numbers that measure the degree of precision, is a minimum."

In conclusion, the principle that the sum of the squares of the differences between the observed and computed quantities must be a minimum, may be considered independently of the calculus of probabilities.

An observation: Other criteria could be used. "But of all these principles ours is the most simple; by the others we should be led into the most complicated calculations."

THEORIA

# MOTVS CORPORVM COELESTIVM

ťN

SECTIONIBUS CONICIS SOLEM AMBIENTIUM

AVCTORE

CAROLO FRIDERICO GAVSS.

HAMOVEGE SYMPHES PRIP. PROPERTY I. H. Bresser 1800.

Mathematical Formulation—Regression Model

$$y(t) = \varphi_1(t)\theta_1 + \varphi_2(t)\theta_2 + \dots + \varphi_n(t)\theta_n = \varphi(t)^T \theta$$

- y-observed data
- $\theta_i$ —unknown parameters
- $\varphi_i$ —known functions regression variables

Some notations

$$\varphi^{T}(t) = [\varphi_{1}(t) \ \varphi_{2}(t) \dots \varphi_{n}(t)], \quad \theta^{T} = [\theta_{1} \ \theta_{2} \dots \theta_{n}] 
Y(t) = [y(1) \ y(2) \dots y(t)]^{T}, \quad E(t) = [\varepsilon(1) \ \varepsilon(2) \dots \varepsilon(t)]^{T} 
\Phi(t) = \begin{pmatrix} \varphi^{T}(1) \\ \vdots \\ \varphi^{T}(t) \end{pmatrix}, \quad P(t) = \left(\sum_{i=1}^{t} \varphi(i)\varphi^{T}(i)\right)^{-1} = \left(\Phi^{T}(t)\Phi(t)\right)^{-1} 
\varepsilon(i) = y(i) - \hat{y}(i) = y(i) - \varphi^{T}(i)\theta$$

# **Solving the LS Problem**

Minimize with respect to  $\theta$ 

$$V( heta,t) = rac{1}{2}\sum_{i=1}^t arepsilon(i)^2 = rac{1}{2}\sum_{i=1}^t ig(y(i) - arphi^T(i) hetaig)^2 = rac{1}{2}\,\, heta^T A heta - b^T heta + rac{1}{2}\,\,c$$

where

$$A = \sum_{i=1}^t arphi(i)arphi^T(i), \quad b \quad = \sum_{i=1}^t arphi(i)y(i), \quad c = \sum_{i=1}^t y^2(i)$$

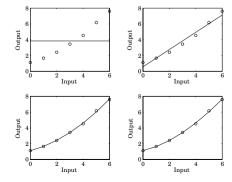
The parameter  $\hat{\theta}$  that minimizes the loss function are given by the **normal equations** 

$$A\hat{ heta} = b, \;\; \Rightarrow \hat{ heta} = A^{-1}b = Pb$$

If the matrix A is nonsingular, the minimum is unique

## **Example Continued**

Model	$\hat{b}_0$	$\hat{b}_1$	$\hat{b}_2$	$\hat{b}_3$	V
1	3.85				34.46
2	0.57	1.09			1.01
3	1.11	0.45	0.11		0.031
4	1.13	0.37	0.14	-0.003	0.027



# An Example

$$y(t) = b_0 + b_1 u(t) + b_2 u^2(t) + e(t)$$

$$\sigma = 0.1$$

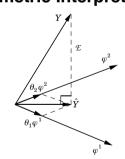
$$\varphi^T(T) = \begin{bmatrix} 1 & u(t) & u^2(t) \end{bmatrix}$$

$$\theta^T = \begin{bmatrix} b_0 & b_1 & b_2 \end{bmatrix}$$

Estimated models

Model 1:  $y(t) = b_0$ Model 2:  $y(t) = b_0 + b_1 u$ Model 3:  $y(t) = b_0 + b_1 u + b_2 u^2$ Model 4:  $y(t) = b_0 + b_1 u + b_2 u^2 + b_3 u^3$ 

**Geometric Interpretation** 



$$E = Y - \varphi^1 \theta_1 - \varphi^2 \theta_2 - \dots - \varphi^n \theta_n$$

Find the smallest possible E?

$$(\varphi^i)^T (y - \theta_1 \varphi^1 - \theta_2 \varphi^2 - \dots - \theta_n \varphi^n) = 0$$

The normal equations!

# **Statistical Interpretation**

$$y(t) = \varphi^{T}(t)\theta^{0} + e(t)$$

 $\theta^0$  – "true" parameters

e(t) – independent random variables with zero mean and variance  $\sigma^2$ 

If  $\Phi^T \Phi$  is nonsingular, then

$$E\hat{ heta}= heta^0$$
 cov  $\hat{ heta}=\sigma^2(\Phi^T\Phi)^{-1}=\sigma^2P$   $s^2=2V(\hat{ heta},t)/(t-n)$ 

is an unbiased estimate of  $\sigma^2$ 

n – number of parameters in  $\theta^0$ 

t – number of data

# **Recursive Least Squares**

The LS estimate is given by

$$\begin{split} \hat{\theta}(t) &= P(t) \left( \sum_{i=1}^{t} \varphi(i) y(i) + \varphi(t) y(t) \right) \\ P(t) &= (\sum_{i=1}^{t} \varphi(i) \varphi^{T}(i))^{-1}, \quad P(t)^{-1} = P(t-1)^{-1} + \varphi(t) \varphi^{T}(t) \end{split}$$

But

$$\sum_{i=1}^{t-1} \varphi(i) y(i) = P(t-1)^{-1} \hat{\theta}(t-1) = P(t)^{-1} \hat{\theta}(t-1) - \varphi(t) \varphi^T(t) \hat{\theta}(t-1)$$

The estimate at time t can now be written as

$$\hat{\theta}(t) = \hat{\theta}(t-1) - P(t)\varphi(t)\varphi^{T}(t)\hat{\theta}(t-1) + P(t)\varphi(t)y(t) 
= \hat{\theta}(t-1) + P(t)\varphi(t)\left(y(t) - \varphi^{T}(t)\hat{\theta}(t-1)\right) = \hat{\theta}(t-1) + K(t)\varepsilon(t)$$

Want recursive equation for P(t) not for  $P(t)^{-1}$ 

# **Recursive Least Squares**

Idea:

- Want to avoid repeating all calculations if data new data arrives recursively
- Does there exist a recursive formula that expresses  $\hat{\theta}(t)$  in terms of  $\hat{\theta}(t-1)$ ?

#### **The Matrix Inversion Lemma**

Let A, C, and  $(C^{-1}+DA^{-1}B)$  be nonsingular square matrices. Then

$$(A + BCD)^{-1}$$
  
=  $A^{-1} - A^{-1}B(C^{-1} + DA^{-1}B)^{-1}DA^{-1}$ 

Prove by direct substitution

Given  $A^{-1}$  we can get the LHS inverse

What about the inverse on the RHS?

# Recursion for P(t)

The matrix inversion lemma gives

$$\begin{split} P(t) &= \left(\sum_{i=1}^{t} \varphi(i) \varphi^{T}(i)\right)^{-1} \\ &= \left(\sum_{i=1}^{t-1} \varphi(i) \varphi^{T}(i) + \varphi(t) \varphi^{T}(t)\right)^{-1} \\ &= \left(P(t-1)^{-1} + \varphi(t) \varphi^{T}(t)\right)^{-1} \\ &= P(t-1) - P(t-1) \varphi(t) \\ &\times \left(I + \varphi^{T}(t) P(t-1) \varphi(t)\right)^{-1} \varphi^{T}(t) P(t-1) \end{split}$$

Hence

$$\begin{split} K(t) &= P(t)\varphi(t) \\ &= P(t-1)\varphi(t) \left(I + \varphi^T(t)P(t-1)\varphi(t)\right)^{-1} \end{split}$$

#### **Example: Recursive estimate of a constant**

Consider the following noisy observation of a constant parameter

$$y_k = \theta + v_k$$
,  $\mathcal{E}\{v_k\} = 0$ ,  $\mathcal{E}\{v_i v_i\} = \sigma^2 \delta_{ij}$ 

Use linear regression of  $y_k = \phi_k \theta + v_k$  with  $\phi_k = 1$  for all k.

A least-squares estimate is found as the sample average

$$\widehat{\theta}_k = \frac{1}{k} \sum_{i=1}^k y_i$$

Include previously made summations in some state  $\widehat{\theta}_k$ , which is updated when new data arrive!!

A feasible recursive state equation

$$\widehat{\theta}_k = \widehat{\theta}_{k-1} + \frac{1}{k} (y_k - \widehat{\theta}_{k-1})$$

$$p_k = \sigma^2 (\sum_{i=1}^k \phi_i \phi_i^T)^{-1} \approx \mathcal{E}\{(\widehat{\theta} - \theta)(\widehat{\theta} - \theta)^T\}$$

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# **Recursive Least-Squares RLS**

$$\begin{split} \hat{\theta}(t) &= \hat{\theta}(t-1) + K(t) \big( y(t) - \varphi^T(t) \hat{\theta}(t-1) \big) \\ K(t) &= P(t) \varphi(t) \\ &= P(t-1) \varphi(t) \big( I + \varphi^T(t) P(t-1) \varphi(t) \big)^{-1} \\ P(t) &= P(t-1) - P(t-1) \varphi(t) \\ &\qquad \times \big( I + \varphi^T(t) P(t-1) \varphi(t) \big)^{-1} \varphi^T(t) P(t-1) \\ &= \big( I - K(t) \varphi^T(t) \big) P(t-1) \end{split}$$

- Intuitive interpretation
- Kalman filter
- Interpretation of  $\theta$  and P
- Initial values  $(P(0) = r \cdot I)$

**Example: Recursive estimate of a constant** The parameter variance estimate can be expressed as a state vector with updating in each recursion according to

$$p_k^{-1} = p_{k-1}^{-1} + \frac{1}{\sigma^2}$$

or

$$p_k = \frac{\sigma^2 p_{k-1}}{\sigma^2 + p_{k-1}}$$

where it can be noticed that  $p_k \to 0$  as  $k \to \infty$ .  $\square$ 

#### Some properties of recursive least-squares estimation

For convergence analysis, consider the following quadratic function of the parameter error

$$Q(\widehat{\theta}_k) = \frac{1}{2}(\widehat{\theta}_k - \theta)^T P_k^{-1}(\widehat{\theta}_k - \theta) = \frac{1}{2}\widetilde{\theta}_k^T P_k^{-1}\widetilde{\theta}_k$$

This function develops in each recursion according to

$$\begin{split} Q(\widehat{\theta}_k) - Q(\widehat{\theta}_{k-1}) &= \frac{1}{2} \widetilde{\theta}_k^T P_k^{-1} \widetilde{\theta}_k - \frac{1}{2} \widetilde{\theta}_{k-1}^T P_{k-1}^{-1} \widetilde{\theta}_{k-1} \\ &= \frac{1}{2} \widetilde{\theta}_{k-1}^T (P_k^{-1} - P_{k-1}^{-1}) \widetilde{\theta}_{k-1} + \widetilde{\theta}_{k-1}^T \phi_k \varepsilon_k + \frac{1}{2} \phi_k^T P_k \phi \\ &= \frac{1}{2} (\widetilde{\theta}_{k-1}^T \phi_k + \varepsilon_k)^2 + \frac{1}{2} (-1 + \phi_k^T P_k \phi_k) \varepsilon_k^2 \\ &= \frac{1}{2} (\widetilde{\theta}_{k-1}^T \phi_k + \varepsilon_k)^2 - \frac{1}{2} \frac{\phi_k^T P_{k-1} \phi_k}{1 + \phi_t^T P_{k-1} \phi_k} \varepsilon_k^2 \end{split}$$

Under the assumption  $y_k = \phi_k^T \theta + v_k$  so that  $\varepsilon_k = -\widetilde{\theta}_{k-1}^T \phi_k + v_k$ 

$$Q(\widehat{\theta}_k) - Q(\widehat{\theta}_{k-1}) = \frac{1}{2}v_k^2 - \frac{1}{2}\frac{\phi_k^T P_{k-1}\phi_k}{1 + \phi_k^T P_{k-1}\phi_k}\varepsilon_k^2$$

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# **Forgetting Factor**

 $0<\lambda \leq 1$ 

Equivalent time constant

$$e^{-h/T}=\lambda,\quad T=-rac{h}{\log\lambda}pproxrac{h}{1-\lambda}$$

Rule of thumb: Memory decay to 10% after

$$N = \frac{2}{1 - \lambda}$$

- + Forgets old information
- + Adapts quickly when the process changes
- - The estimates get noise sensitive
- — The *P* matrix may grow (Wind-up)

# **Time-varying Parameters**

Loss function with discounting

$$V( heta,t) = rac{1}{2} \sum_{i=1}^t \lambda^{t-i} \left( y(i) - arphi^T(i) heta 
ight)^2$$

The LS estimate then becomes

$$\begin{split} \hat{\theta}(t) &= \hat{\theta}(t-1) + K(t) \big( y(t) - \varphi^T(t) \hat{\theta}(t-1) \big) \\ K(t) &= P(t) \varphi(t) \\ &= P(t-1) \varphi(t) \left( \lambda I + \varphi^T(t) P(t-1) \varphi(t) \right)^{-1} \\ P(t) &= \left( I - K(t) \varphi^T(t) \right) P(t-1) / \lambda \end{split}$$

## **Continuous Time Models**

Regression model

$$y(t) = \varphi_1(t)\theta_1 + \varphi_2(t)\theta_2 + \dots + \varphi_n(t)\theta_n = \varphi(t)^T \theta$$

Loss function with forgetting

$$V( heta) = \int_0^t e^{lpha(t-s)} \left(y(s) - arphi^T(s) heta
ight)^2 ds$$

The normal equations

$$\int_0^t e^{-lpha(t-s)} arphi(s) arphi^T(s) ds \hat{ heta}(t) = \int_0^t e^{-lpha(t-s)} arphi(s) y(s) ds$$

Estimate is unique if the matrix

$$R(t) = \int_0^t e^{-lpha(t-s)} arphi(s) arphi^T(s) \, ds$$

is positive definite.

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# **Recursive Equations for Continuous Time Models**

Regression model

$$y(t) = \varphi(t)^T \theta$$

Recursive least equations

$$\begin{split} \frac{d\hat{\theta}}{dt} &= P(t)\varphi(t)e(t) \\ e(t) &= y(t) - \varphi^T(t)\hat{\theta}(t) \\ \frac{dP(t)}{dt} &= \alpha P(t) - P(t)\varphi(t)\varphi^T(t)P(t) \\ \frac{dR}{dt} &= -\alpha R + \varphi \varphi^T \end{split}$$

# **Estimating Parameters in Dynamical Systems**

Basic idea: Rewrite the equations as a regression model!

- · Dynamical systems
  - FIR models
  - ARMA models
  - Continuous time models
  - Nonlinear models
- Experimental conditions
  - Excitation
  - Closed loop identification

## **Real-time Parameter Estimation**

- 1. Introduction
- 2. Least squares and regression
- 3. Dynamical systems
- 4. Experimental conditions
- 5. Examples
- 6. Conclusions

# Finite Impulse Response (FIR) Models

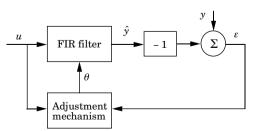
$$y(t) = b_1 u(t-1) + b_2 u(t-2) + \dots + b_n u(t-n)$$

or

$$y(t) = \varphi^T(t-1)\theta, \quad \theta = [b_1 \dots b_n]^T$$
  
 $\varphi^T(t-1) = [u(t-1) \dots u(t-n)]$ 

A regression model!

$$\hat{y}(t) = \hat{b}_1(t-1)u(t-1) + \cdots + \hat{b}_n(t-1)u(t-n)$$



## **Pulse Transfer Function Models**

$$y(t) + a_1 y(t-1) + \dots + a_n y(t-n) =$$
  
 $b_1 u(t-1) + \dots + b_n u(t-n)$ 

Write as

$$y(t) = \varphi(t-1)^T \theta$$

where

$$\varphi(t-1)$$

$$= [-y(t-1)\dots - y(t-n)u(t-1)\dots u(t-n)]^T$$

$$\theta = [a_1 \dots a_n b_1 \dots b_n]^T$$

- Autoregression!
- Equation error

# **Nonlinear Models**

Consider the model

$$y(t) + ay(t-1) = b_1u(t-1) + b_2u^2(t-1)$$

Introduce

$$\theta = \left[ a \ b_1 \ b_2 \right]^T$$

and

$$\varphi^{T}(t) = \left[ -y(t) \ u(t) \ u^{2}(t) \right]$$

Hence

$$y(t) = \varphi^T(t-1)\theta$$

Autoregression!

Linearity in the parameters

#### **Transfer Function Models**

$$rac{d^ny}{dt^n}+a_1rac{d^{n-1}y}{dt^{n-1}}+\cdots+a_ny=b_1rac{d^{n-1}u}{dt^{n-1}}+\cdots+b_nu$$

as

$$A(p)y(t) = B(p)u(t)$$

Introduce

$$A(p)y_f(t) = B(p)u_f(t), \quad y_f(t) = F(p)y(t), \quad u_f(t) = F(p)u(t)$$

and F(p) has pole excess greater than n

$$\theta = \begin{bmatrix} a_1 \dots a_n & b_1 \dots b_n \end{bmatrix}^T$$

$$\varphi_f(t) = \begin{bmatrix} -p^{n-1} y_f \dots - y_f & p^{n-1} u_f \dots u_f \end{bmatrix}$$

$$= \begin{bmatrix} -p^{n-1} F(p) y \dots - F(p) y & p^{n-1} F(p) u \dots F(p) u \end{bmatrix}$$

Hence, a regression model  $y_f(t) = \varphi_f^T(t)\theta$ 

# **Experimental Conditions**

- Excitation
- · Closed loop identification
- Model structure

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#### **Persistent Excitation**

The matrix  $\sum_{k=n+1}^{t} \varphi(k) \varphi^{T}(k)$  is given by

$$\begin{bmatrix} \sum_{n+1}^{t} u^{2}(k-1) & \dots & \sum_{n+1}^{t} u(k-1)u(k-n) \\ \vdots & & & \\ \sum_{n+1}^{t} u(k-1)u(k-n) & \dots & \sum_{n+1}^{t} u^{2}(k-n) \end{bmatrix}$$

$$C_{n} = \lim_{t \to \infty} \frac{1}{t} \Phi^{T} \Phi, \quad C_{n} = \begin{bmatrix} c(0) & c(1) & \dots & c(n-1) \\ c(1) & c(0) & \dots & c(n-2) \\ \vdots & & & \\ c(n-1) & c(n-2) & \dots & c(0) \end{bmatrix}$$

$$c(k) = \lim_{t \to \infty} \frac{1}{t} \sum_{i=1}^{t} u(i)u(i-k)$$

A signal u is called *persistently exciting* (PE) of order n if the <sup>32</sup> matrix C is positive definite.

# **Examples**

A signal u is called *persistently exciting* (PE) of order n if the matrix  $C_n$  is positive definite. An equivalent condition

· A step is PE of order 1

$$(q-1)u(t) = 0$$

· A sinusoid is PE of order 2

$$(q^2 - 2q\cos\omega h + 1)u(t) = 0$$

- · White noise
- PRBS
- · Physical meaning
- · Mathematical meaning

#### **Another Characterization**

A signal u is persistently exciting of order n if and only if

$$U=\lim_{t o\infty}rac{1}{t}\sum_{k=1}^t (A(q)u(k))^2>0$$

for all nonzero polynomials A of degree n-1 or less.

**Proof** Let the polynomial A be

$$A(q) = a_0 q^{n-1} + a_1 q^{n-2} + \dots + a_{n-1}$$

A straightforward calculation gives

$$egin{aligned} U &= \lim_{t o \infty} rac{1}{t} \sum_{k=1}^t \left( a_0 u(k+n-1) + \cdots + a_{n-1} u(k) 
ight)^2 \ &= a^T C_n a \end{aligned}$$

# Loss of Identifiability due to Feedback

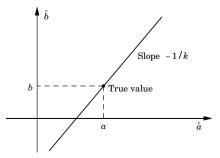
$$y(t) = ay(t-1) + bu(t-1) + e(t), \quad u(t) = -ky(t)$$

Multiply by  $\alpha$  and add, then

$$y(t) = (a + \alpha k)y(t - 1) + (b + \alpha)u(t - 1) + e(t)$$

Same I/O relation for all  $\hat{a}$  and  $\hat{b}$  such that

$$\hat{a} = a + \alpha k, \quad \hat{b} = b + \alpha$$



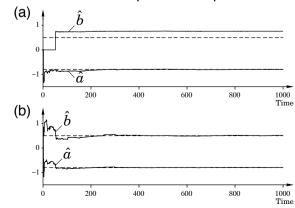
## **Real-time Parameter Estimation**

- 1. Introduction
- 2. Least squares and regression
- 3. Dynamical systems
- 4. Experimental conditions
- 5. Examples
- 6. Conclusions

# **Excitation**

## Input:

- Unit pulse at t = 50
- Square wave of unit amplitude and period 100



# **Examples**

Model

$$y(t) + ay(t-1) = bu(t-1) + e(t)$$

**Parameters** 

$$a = -0.9$$

$$b = 0.5$$

$$\sigma = 0.5$$

$$\hat{\theta}(0) = 0$$

$$P(0) = 100$$

$$\hat{\theta} = \begin{pmatrix} \hat{a} \\ \hat{b} \end{pmatrix}$$

$$\varphi(t-1) = (-y(t-1) \quad u(t-1))$$

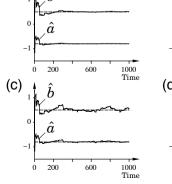
# **Forgetting Factor**

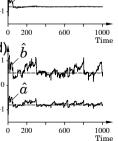
Recall

(a)

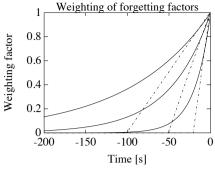
$$Tpprox rac{h}{1-\lambda}, \quad N=rac{2}{1-\lambda}$$

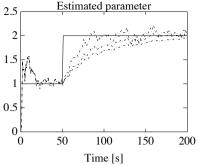
Parameters:  $\lambda = 1$ ,  $\lambda = 0,999$ ,  $\lambda = 0.99$ ,  $\lambda = 0.95$ 

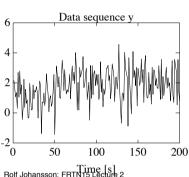


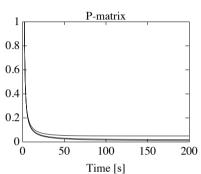


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# **Conclusions**

What you should remember:

- The least squares method
- The normal equations
- The recursive equations
- · The matrix inversion lemma

What you should master:

- The recursive equations
- · The role of excitation

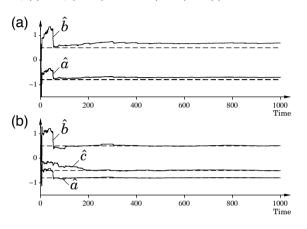
Role in adaptive control

- Recursive estimation is a key part of adaptive control
- · Recursive least squares is a useful method

## **Colored Noise**

Process model and Estimator model

$$y(t) - 0.8y(t-1) = 0.5u(t-1) + e(t) - 0.5e(t-1)$$
  
$$y(t) + ay(t-1) = bu(t-1) + e(t)$$



# **Innovations Model**

It is also possible to represent the output process y with only one noise source w.

One can find K so that y can be represented as

$$\hat{x}_{k+1} = A\hat{x}_k + Kw_k 
y_k = C\hat{x}_k + w_k$$

(This is obtained from the optimal Kalman predictor in Ch7)