

## FRTN10 Multivariable Control, Lecture 11

Automatic Control LTH, 2016

### Course Outline

- L1-L5 Specifications, models and loop-shaping by hand
- L6-L8 Limitations on achievable performance
- L9-L11 Controller optimization: Analytic approach
  - 9. Linear-quadratic optimal control
  - 10. Optimal observer-based feedback
  - 11. **More on LQG**
- L12-L14 Controller optimization: Numerical approach

### Lecture 11 – Outline

1. Tuning the LQG design parameters
2. Robustness of LQG
3. Integral action, reference values

[Glad&Ljung sections 9.1–9.4 and 5.7]

### Recall the main result of LQG

Given white noise  $(v_1, v_2)$  with intensity  $R$  and the linear plant

$$\begin{cases} \dot{x}(t) = Ax(t) + Bu(t) + Nv_1(t) \\ y(t) = Cx(t) + v_2(t) \end{cases} \quad R = \begin{bmatrix} R_1 & R_{12} \\ R_{12}^T & R_2 \end{bmatrix}$$

consider controllers of the form  $u = -L\hat{x}$  with  $\frac{d}{dt}\hat{x} = A\hat{x} + Bu + K[y - C\hat{x}]$ . The stationary variance

$$\mathbf{E} \left( x^T Q_1 x + 2x^T Q_{12} u + u^T Q_2 u \right)$$

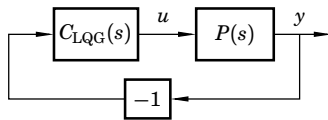
is minimized when

$$\begin{aligned} K &= (PC^T + NR_{12})R_2^{-1} \quad L = Q_2^{-1}(SB + Q_{12})^T \\ 0 &= Q_1 + A^T S + SA - (SB + Q_{12})Q_2^{-1}(SB + Q_{12})^T \\ 0 &= NR_1 N^T + AP + PA^T - (PC^T + NR_{12})R_2^{-1}(PC^T + NR_{12})^T \end{aligned}$$

The minimal variance is

$$\text{tr}(SNR_1 N^T) + \text{tr}[PL^T(B^T S B + Q_2)L]$$

### The LQG controller



The controller transfer function (from  $-y$  to  $u$ ) is given by

$$C_{LQG}(s) = L(sI - A + BL + KC)^{-1}K$$

- Same order as the (extended) plant model
- Strictly proper (if  $R_2 > 0$ )

Matlab: `reg(P,L,K)` or `lqgreg(Kest,L)`

### Lecture 11 – Outline

Tuning the LQG design parameters

Robustness of LQG

Integral action, reference values

### How to choose the cost function

- In rare instances, a quadratic cost function follows directly from the design specifications
- In most cases,  $Q_1$ ,  $Q_2$ ,  $Q_{12}$  must be tuned by the designer to achieve the desired closed-loop behavior

Starting point:

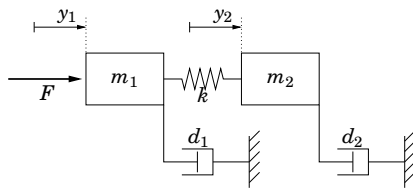
- Put  $Q_{12} = 0$  and make  $Q_1$ ,  $Q_2$  diagonal
- Make the diagonal elements equal to the inverse value of the square of the allowed deviation:

$$\begin{aligned} & x(t)^T Q_1 x(t) + u(t)^T Q_2 u(t) \\ &= \left( \frac{x_1(t)}{x_1^{\max}} \right)^2 + \dots + \left( \frac{x_n(t)}{x_n^{\max}} \right)^2 + \left( \frac{u_1(t)}{u_1^{\max}} \right)^2 + \dots + \left( \frac{u_m(t)}{u_m^{\max}} \right)^2 \end{aligned}$$

### How to tune the cost function

- To make the controller more aggressive, decrease  $Q_2$  (or, equivalently, increase  $Q_1$ )
- To increase the damping of a state  $x_i$ , add penalty on  $\dot{x}_i^2$  (may give cross-terms)
- To make a state  $x_i$  behave more like  $\dot{x}_i = -\alpha x_i$ , add penalty on  $(\dot{x}_i + \alpha x_i)^2$  (gives cross-terms)

## Example — Flexible servo

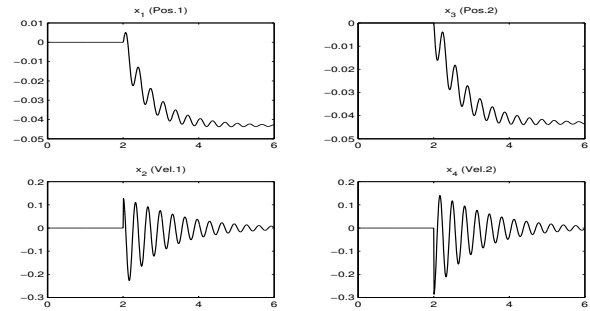


$$m_1 \frac{d^2 y_1}{dt^2} = -d_1 \frac{dy_1}{dt} - k(y_1 - y_2) + F(t)$$

$$m_2 \frac{d^2 y_2}{dt^2} = -d_2 \frac{dy_2}{dt} + k(y_1 - y_2)$$

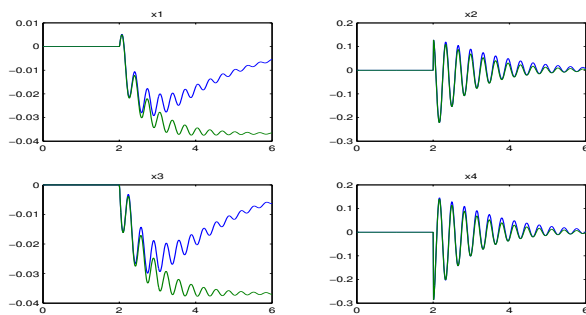
Introduce state variables  $x_1 = y_1$ ,  $x_2 = \dot{y}_1$ ,  $x_3 = y_2$ ,  $x_4 = \dot{y}_2$

## Open loop response



## Penalize velocity error or position error?

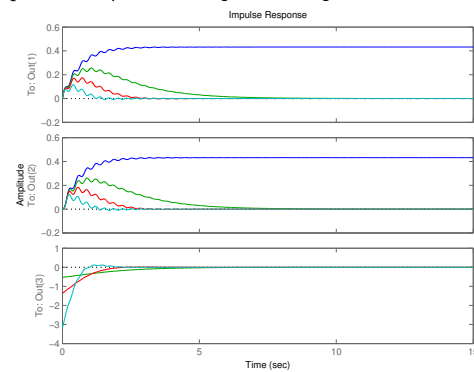
Minimize  $\mathbf{E}[x_2(t)^2 + x_4(t)^2 + u(t)^2]$  or  $\mathbf{E}[x_1(t)^2 + x_3(t)^2 + u(t)^2]$  ?



When only velocity is penalized, a static position error remains

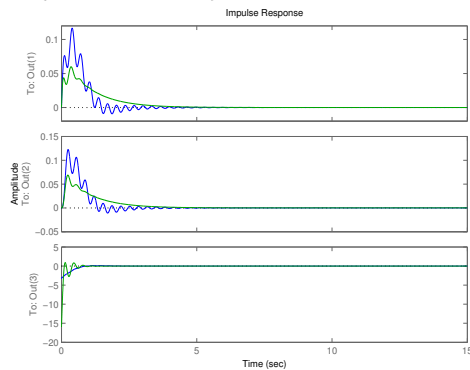
## Position error control

Response of  $x_1(t), x_3(t), u(t) = -Lx(t)$  to impulse disturbance.  
 $Q_1 = \text{diag}\{q, 0, q, 0\}$  ( $q = 0, 1, 10, 100$ ),  $Q_{12} = 0$ ,  $Q_2 = 1$ .  
 Large  $q \Rightarrow$  fast response but large control signal.



## Position+velocity error control

To reduce oscillations, penalize also velocity error. Comparison between  $Q_1 = \text{diag}\{100, 0, 100, 0\}$  and  $Q_1 = \text{diag}\{100, 100, 100, 100\}$ .



## How to choose/tune the noise intensities

- If no accurate noise characterization is available, as a starting point put  $N = B$ ,  $R_1 = I$ , and  $R_2 = \rho I$ , where  $\rho$  is a scalar
- To make the controller less sensitive to measurement noise, increase  $R_2$  (also makes the controller less aggressive)

## Lecture 11 – Outline

Tuning the LQG design parameters

Robustness of LQG

Integral action, reference values

## Robustness of LQG controllers

Candidate for the best abstract ever:

### Guaranteed Margins for LQG Regulators

JOHN C. DOYLE

**Abstract—**There are none.

### INTRODUCTION

Considerable attention has been given lately to the issue of robustness of linear-quadratic (LQ) regulators. The recent work by Safonov and Athans [1] has extended to the multivariable case the now well-known guarantee of 60° phase and 6 dB gain margin for such controllers. However, for even the single-input, single-output case there has remained the question of whether there exist any guaranteed margins for the full LQG (Kalman filter in the loop) regulator. By counterexample, this note answers that question; there are none.

[IEEE Transactions on Automatic Control, 23:4, 1978]

### Example (Doyle–Stein, 1979)

Nice minimum-phase SISO plant, no fundamental limitations:

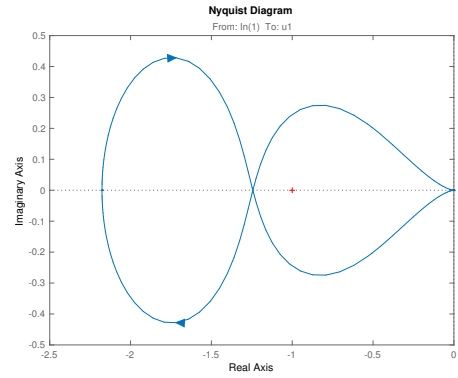
$$A = \begin{bmatrix} -4 & -3 \\ 1 & 0 \end{bmatrix}, \quad B = \begin{bmatrix} 1 \\ 0 \end{bmatrix}, \quad N = \begin{bmatrix} -61 \\ 35 \end{bmatrix}, \quad C = \begin{bmatrix} 1 & 2 \end{bmatrix}$$

$$Q_1 = 80 \begin{bmatrix} 1 & \sqrt{35} \\ \sqrt{35} & 35 \end{bmatrix}, \quad Q_2 = 1, \quad R_1 = 1, \quad R_2 = 1$$

gives

- ▶ Control poles:  $-7 \pm 2i$
- ▶ Observer poles:  $-7.02 \pm 1.95i$

### Example (Doyle–Stein, 1979)



$$M_s = 4.8, \quad \phi_m = 14.8^\circ$$

### Loop transfer recovery

The robustness of an LQG controller can often be improved by either

- ▶ adding a penalty proportional to  $C^T C$  to  $Q_1$
- ▶ adding a penalty proportional to  $B B^T$  to  $N R_1 N^T$

Makes the loop transfer function more similar to the state feedback (LQ) loop gain

Price: Higher controller gain

### Lecture 11 – Outline

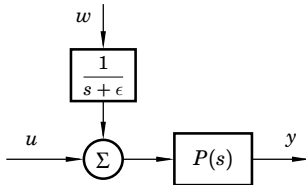
Tuning the LQG design parameters

Robustness of LQG

Integral action, reference values

### Integral action via disturbance modeling

Extend the plant model with a low-frequency disturbance acting on the process input:



The Kalman filter will include a (near) integrator

Easy to generalize to other disturbance models

### Integral action via explicit integration

Add explicit integrators on the plant outputs

Gives extended plant model

$$\begin{bmatrix} \dot{x} \\ \dot{x}_i \end{bmatrix} = \begin{bmatrix} A & 0 \\ C & 0 \end{bmatrix} \begin{bmatrix} x \\ x_i \end{bmatrix} + \begin{bmatrix} B \\ 0 \end{bmatrix} u + \begin{bmatrix} N \\ 0 \end{bmatrix} v_1$$

Extended state feedback law from LQ design:

$$u = - \begin{bmatrix} L & L_i \end{bmatrix} \begin{bmatrix} x \\ x_i \end{bmatrix}$$

Including a penalty on  $x_i$  in the LQ design makes  $x \rightarrow 0$  in case of a constant input load disturbance

### Handling reference values

Simple solution:

$$u(t) = -L\hat{x}(t) + L_r r(t)$$

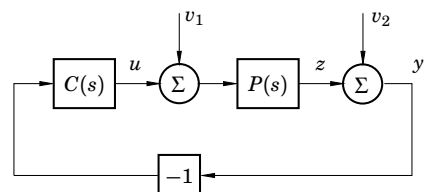
Assuming  $z = Mx$  and  $\dim z = \dim u$ , select

$$L_r = [M(BL - A)^{-1}B]^{-1}$$

to ensure static gain  $I$  from  $r$  to  $z$  (in absence of disturbances)

A reference filter to further shape  $G_{yr}(s)$  can be added if needed

### LQG example — DC-servo



$$\text{Process: } P(s) = \frac{20}{s(s+1)}$$

$$\text{Cost function: } J = \mathbf{E} (z^2 + u^2)$$

$$\text{White noise intensities: } R_1 = 1, R_2 = 1, R_{12} = 0$$

## LQG design

State-space model:

$$\begin{bmatrix} \dot{x}_1 \\ \dot{x}_2 \end{bmatrix} = \underbrace{\begin{bmatrix} 0 & 0 \\ 1 & -1 \end{bmatrix}}_A \begin{bmatrix} x_1 \\ x_2 \end{bmatrix} + \underbrace{\begin{bmatrix} 20 \\ 0 \end{bmatrix}}_B u + \underbrace{\begin{bmatrix} 20 \\ 0 \end{bmatrix}}_N v_1$$

$$y = x_2 + v_2 \quad z = x_2$$

Cost matrices:

$$Q_1 = \begin{bmatrix} 0 & 0 \\ 0 & 1 \end{bmatrix} \quad Q_2 = 1$$

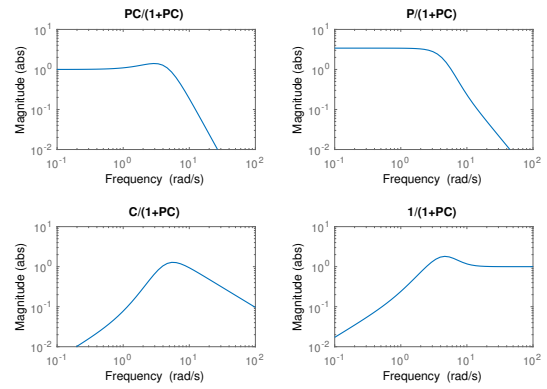
Solving the Riccati equations gives the optimal controller

$$\frac{d}{dt} \hat{x} = (A - BL)\hat{x} + K[y - C\hat{x}] \quad u = -L\hat{x}$$

where

$$L = \begin{bmatrix} 0.2702 & 0.7298 \end{bmatrix} \quad K = \begin{bmatrix} 20.0000 \\ 5.4031 \end{bmatrix}$$

## Gang of four



Nonzero static gain in  $\frac{P}{1+PC}$  indicates poor disturbance rejection

## Integral action

Add explicit integrator on the output and extend the model:

$$\begin{bmatrix} \dot{x}_1 \\ \dot{x}_2 \\ \dot{x}_i \end{bmatrix} = \underbrace{\begin{bmatrix} 0 & 0 & 0 \\ 1 & -1 & 0 \\ 0 & 1 & 0 \end{bmatrix}}_{A_e} \begin{bmatrix} x_1 \\ x_2 \\ x_i \end{bmatrix} + \underbrace{\begin{bmatrix} 20 \\ 0 \\ 0 \end{bmatrix}}_{B_e} u + \underbrace{\begin{bmatrix} 20 \\ 0 \\ 0 \end{bmatrix}}_{N_e} v_1$$

$$y = x_2 + v_2, \quad z_1 = x_2, \quad z_2 = x_i$$

Minimization of  $x_2^2 + 0.01x_i^2 + u^2$  gives the optimal state feedback

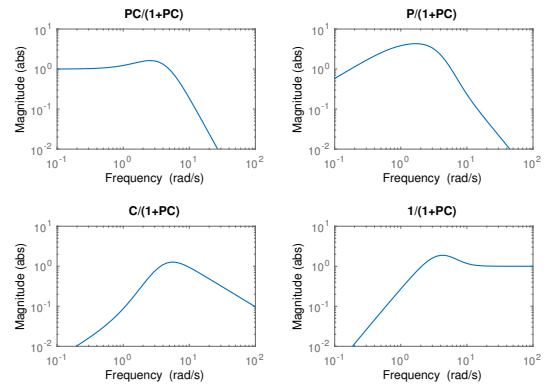
$$u = -L_e \begin{bmatrix} \hat{x} \\ x_i \end{bmatrix}$$

where

$$L_e = \begin{bmatrix} 0.2751 & 0.7569 & -0.1 \end{bmatrix}$$

We use the same Kalman filter as before

## Gang of four with integral action



## Summary of LQG

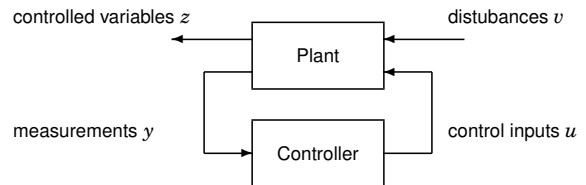
### Advantages

- Works fine with multivariable models
- Observer structure ties to reality
- Always stabilizing
- Well developed theory, analytic solutions

### Disadvantages

- High-order controllers (same order as the extended plant model)
- Sometimes hard to choose weights
- No robustness guarantees – must always check the resulting controller!

## Alternative norms for optimization



$H_\infty$  optimal control:

$$\text{Minimize} \quad \max_{\omega} \|G_{zu}(i\omega)\|$$

Can be solved using a couple of Riccati equations, similar to the LQG problem