

## Lecture 14: Controller simplification

- **Model reduction by balanced truncation**
  - Application to controller simplification
  - Frequency weighted balanced truncation

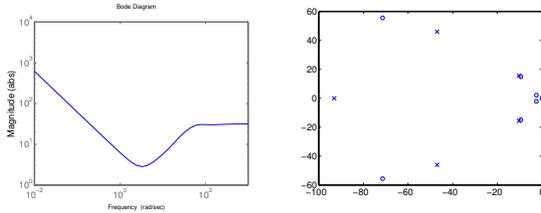
Model reduction by balanced truncation is described in Glad/Ljung, section 3.6.

### The optimized controller has high order

Recall that  $C(s) = [I - Q(s)P_{yu}(s)]^{-1}Q(s)$  with  $Q(s) = \sum_{k=0}^N Q_k \phi_k(s)$ .

Hence the controller order will grow with the number of  $\phi_k$ , or the number of time samples in the impulse response representation.

However, in the DC-servo example, both the Bode diagram and pole-zero diagram of the controller indicate that cancellations can be done to simplify the controller.



### Gramians, looking back

$$\begin{aligned} \frac{dx}{dt} &= Ax + Bu \\ y &= Cx + Du \end{aligned} \quad x(t) = e^{At}x(0) + \int_0^t e^{A(t-\tau)}Bu(\tau)d\tau$$

Impulse response from zero initial condition:  $u_i(t) = \delta(t)$ ,  $x_0 = 0$

$$\begin{aligned} x_i(t) &= e^{At}B_i \\ X(t) &= [x_1 \quad x_2 \quad \dots \quad x_n] = e^{At}B \\ S_x &\hat{=} \int_0^\infty X(t)X^T(t)dt = \int_0^\infty e^{At}BB^T e^{A^T t} dt \end{aligned}$$

Output from  $u = 0$  (only initial condition  $x_0$ )

$$\begin{aligned} y(t) &= Cx(t) = Ce^{At}x_0 \\ \int_0^\infty y(t)^T y(t) dt &= \int_0^\infty x_0^T e^{A^T t} C^T C A t x_0 dt \hat{=} x_0^T O_x x_0 \end{aligned}$$

### Hankel singular values

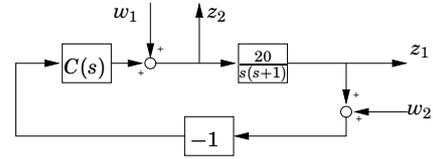
Notice that

$$\begin{bmatrix} \sigma_1^2 & & 0 \\ & \ddots & \\ 0 & & \sigma_n^2 \end{bmatrix} = \underbrace{(TS_x T^T)}_{\Sigma} \underbrace{(T^{-T} O_x T^{-1})}_{\Sigma} = TS_x O_x T^{-1}$$

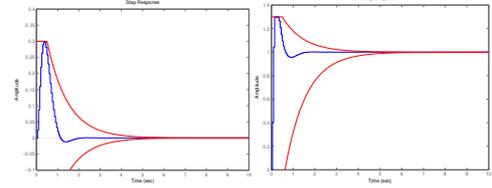
so the diagonal elements are the eigenvalues of  $S_x O_x$ , independently of coordinate system. The numbers  $\sigma_1, \dots, \sigma_n$  are called the *Hankel singular values* of the system.

A small Hankel singular value corresponds to a state that is both weakly controllable and weakly observable. Hence, it can be truncated without much effect on the input-output behavior.

## Example — DC-motor



We previously minimized  $\int_{-\infty}^{\infty} G_{zw}(i\omega)G_{zw}(i\omega)^* d\omega$  subject to step response bounds on the transfer functions from  $w_1$  and  $w_2$  to  $z_1$ :



### Controllability and Observability

The controllability Gramian  $S = \int_0^\infty e^{At}BB^T e^{A^T t} dt$  can be computed by solving the linear system of equations

$$AS + SA^T + BB^T = 0$$

The observability Gramian  $O = \int_0^\infty e^{A^T t}C^T C e^{At} dt$  can be computed by solving the linear system of equations

$$A^T O + OA + C^T C = 0$$

We want to remove states that are either poorly controllable or poorly observable.

### Balanced Realizations

For a stable system  $(A, B, C)$  with gramians  $S_x$  and  $O_x$ , the variable transformation  $\xi = Tx$  gives the new state matrices  $\hat{A} = TAT^{-1}$ ,  $\hat{B} = TB$ ,  $\hat{C} = CT^{-1}$  and the new gramians

$$\begin{aligned} S_\xi &= \int_0^\infty e^{\hat{A}t} \hat{B} \hat{B}^T e^{\hat{A}^T t} dt = \int_0^\infty T e^{At} B B^T e^{A^T t} T^T dt = TS_x T^T \\ O_\xi &= \int_0^\infty e^{\hat{A}^T t} \hat{C}^T \hat{C} e^{\hat{A}t} dt = \int_0^\infty T^{-T} e^{A^T t} C^T C e^{At} T^{-1} dt = T^{-T} O_x T^{-1} \end{aligned}$$

A particular choice of  $T$  gives  $S_\xi = O_\xi = \underbrace{\begin{bmatrix} \sigma_1 & & 0 \\ & \ddots & \\ 0 & & \sigma_n \end{bmatrix}}_{\Sigma}$

The corresponding realization

$$\begin{cases} \dot{\xi} = \hat{A}\xi + \hat{B}u \\ y = \hat{C}\xi \end{cases}$$

is called a *balanced realization*.

### Model reduction by balanced truncation

Consider a balanced realization

$$\begin{aligned} \begin{bmatrix} \dot{\xi}_1 \\ \dot{\xi}_2 \end{bmatrix} &= \begin{bmatrix} A_{11} & A_{12} \\ A_{21} & A_{22} \end{bmatrix} \begin{bmatrix} \xi_1 \\ \xi_2 \end{bmatrix} + \begin{bmatrix} B_1 \\ B_2 \end{bmatrix} u & \quad \Sigma = \begin{bmatrix} \Sigma_1 & 0 \\ 0 & \Sigma_2 \end{bmatrix} \\ y &= [C_1 \quad C_2] \begin{bmatrix} \xi_1 \\ \xi_2 \end{bmatrix} + Du \end{aligned}$$

with the lower part of the gramian being  $\Sigma_2 = \begin{bmatrix} \sigma_{r+1} & & 0 \\ & \ddots & \\ 0 & & \sigma_n \end{bmatrix}$ .

Replacing the second state equation by  $\dot{\xi}_2 = 0$  gives the relation  $0 = A_{21}\xi_1 + A_{22}\xi_2 + B_2 u$ . The reduced system

$$\begin{cases} \dot{\xi}_1 = (A_{11} - A_{12}A_{22}^{-1}A_{21})\xi_1 + (B_1 - A_{12}A_{22}^{-1}B_2)u \\ y_r = (C_1 - C_2A_{22}^{-1}A_{21})\xi_1 + (D - C_2A_{22}^{-1}B_2)u \end{cases}$$

satisfies the error bound

$$\frac{\|y - y_r\|_2}{\|u\|_2} \leq 2\sigma_{r+1} + \dots + 2\sigma_n$$

## Example 1

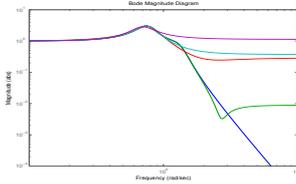
Original system:  $\frac{1-s}{s^6 + 3s^5 + 5s^4 + 7s^3 + 5s^2 + 3s + 1}$

Hankel singular values:

$$\text{Sigma} = [1.9837 \quad 1.9184 \quad 0.7512 \quad 0.3292 \quad 0.1478 \quad 0.0045]$$

Reduced system:

$$\frac{0.3717 s^3 - 0.9682 s^2 + 1.14 s - 0.5185}{s^3 + 1.136 s^2 + 0.825 s + 0.5185}$$



## Example 2 — Heat Exchanger

A state transformation  $\xi_1 = -7.07(x_1 + x_2)$ ,  $\xi_2 = 7.07(x_1 - x_2)$  gives the balanced realization

$$\begin{aligned} \dot{\xi} &= \begin{bmatrix} -0.01 & 0 \\ 0 & -0.41 \end{bmatrix} \xi + 0.0707 \begin{bmatrix} -1 & -1 \\ 1 & -1 \end{bmatrix} u \\ y &= 0.0707 \begin{bmatrix} -1 & 1 \\ -1 & -1 \end{bmatrix} \xi \end{aligned}$$

the common controllability/observability matrix

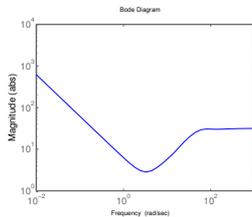
$$S_{\xi} = O_{\xi} = \begin{bmatrix} 0.5 & 0 \\ 0 & 0.0122 \end{bmatrix}$$

and the reduced model

$$\begin{aligned} \dot{\xi}_1 &= -0.01\xi_1 - 0.0707 [1 \quad 1] u \\ y &= -0.0707 \begin{bmatrix} 1 \\ 1 \end{bmatrix} \xi_1 + 0.0122 \begin{bmatrix} 1 & -1 \\ -1 & 1 \end{bmatrix} u \end{aligned}$$

## Reducing the DC-servo Controller

Recall the Bode plot of the optimized controller  $C_{\text{opt}}(s)$ :



The Hankel singular values of  $C_{\text{stab}}(s) = C_{\text{opt}}(s) + \frac{6.17}{s}$  are

$$\text{Sigma} = [16.0768 \quad 2.2306 \quad 0.7023 \quad 0.1994 \quad 0.0896]$$

How many states need to be kept in  $C_{\text{stab}}(s)$ ?

What kind of controller remains?

## Frequency weighted balanced truncation

For model reduction with the aim to minimize

$$\max_{\omega} \|W_o(i\omega)[G(i\omega) - G_r(i\omega)]W_i(i\omega)\|$$

where

$$W_i(s) = C_i(sI - A_i)^{-1}B_i + D_i \quad G(s) = C(sI - A)^{-1}B + D \quad W_o(s) = C_o(sI - A_o)^{-1}B_o + D_o$$

find extended gramians by solving

$$\begin{aligned} \begin{bmatrix} A & BC_i \\ 0 & A_i \end{bmatrix} \begin{bmatrix} S & S_{12} \\ S_{12}^T & S_{22} \end{bmatrix} + \begin{bmatrix} S & S_{12} \\ S_{12}^T & S_{22} \end{bmatrix} \begin{bmatrix} A & BC_i \\ 0 & A_i \end{bmatrix}^T + \begin{bmatrix} BD_i \\ B_i \end{bmatrix} \begin{bmatrix} BD_i \\ B_i \end{bmatrix}^T &= 0 \\ \begin{bmatrix} A & 0 \\ B_o C & A_o \end{bmatrix}^T \begin{bmatrix} O & O_{12} \\ O_{12}^T & O_{22} \end{bmatrix} + \begin{bmatrix} O & O_{12} \\ O_{12}^T & O_{22} \end{bmatrix} \begin{bmatrix} A & 0 \\ B_o C & A_o \end{bmatrix}^T + \begin{bmatrix} C^T D_o^T \\ D_o^T \end{bmatrix} \begin{bmatrix} D_o C & D_o \end{bmatrix} &= 0 \end{aligned}$$

then change coordinates to make  $S$  and  $O$  equal and diagonal before truncating the realization of  $G(s)$  to get  $G_r(s)$  as before.

## Example 2 — Heat Exchanger

$$V_C \frac{dT_C}{dt} = f_C(T_{C_i} - T_C) + \beta(T_H - T_C) \quad (\text{cold side})$$

$$V_H \frac{dT_H}{dt} = f_H(T_{H_i} - T_H) - \beta(T_H - T_C) \quad (\text{hot side})$$

$u_1 = T_{C_i}$  is the in-flow temperature on the cold side  
 $x_1 = T_C$  is the out-flow temperature on the cold side  
 $u_2 = T_{H_i}$  is the in-flow temperature on the hot side  
 $x_2 = T_H$  is the out-flow temperature on the hot side

Numerical values:

$$\begin{aligned} \dot{x} &= \begin{bmatrix} -0.21 & 0.2 \\ 0.2 & -0.21 \end{bmatrix} x + \begin{bmatrix} 0.01 & 0 \\ 0 & 0.01 \end{bmatrix} u \\ y &= x \end{aligned}$$

## DC-servo again

To simplify the controller, we would like to remove states that have little influence on the input-output relationship, i.e. states that are poorly controllable or poorly observable.

For this, we will compute the controllability gramian and the observability gramian. However, these are defined only for stable systems. Hence the integrator needs to be treated separately:

$$C_{\text{opt}}(s) = C_{\text{stab}}(s) - \frac{6.17}{s}$$

For  $C_{\text{stab}}(s)$  the gramians have eigenvalues

$$\begin{aligned} \text{eig}(\text{ConGram}) &= [0.0933 \quad 0.2972 \quad 0.9417 \quad 5.9373 \quad 50.0472] \\ \text{eig}(\text{ObsGram}) &= [0.0291 \quad 0.0913 \quad 0.2964 \quad 1.8811 \quad 17.6379] \end{aligned}$$

Three out of five states are poorly controllable and three are weakly observable. This can be used for reduction!

## Are all frequencies equally important?

The error bound

$$\max_{\omega} |G(i\omega) - G_r(i\omega)| = \sup_u \frac{\|y - y_r\|_2}{\|u\|_2} \leq 2\sigma_{r+1} + \dots + 2\sigma_n$$

emphasizes all frequencies equally, but comparing a controller  $C(s)$  with a reduced controller  $C_r(s)$  in closed loop operation gives

$$|P(I + CP)^{-1}C - P(I + C_r P)^{-1}C_r| \approx |P(I + CP)^{-1}(C - C_r)|$$

Hence it is interesting to minimize the frequency weighted error

$$\max_{\omega} |W(i\omega)[C(i\omega) - C_r(i\omega)]|$$

where  $W(i\omega) = P(i\omega)(I + C(i\omega)P(i\omega))^{-1}$ .

## Summary

- ▶ Low order controllers could be desirable to meet constraints on speed and memory.
- ▶ Balanced realizations can reveal less important states
- ▶ Good theoretical error bounds
- ▶ Frequency weighting essential for closed loop performance
- ▶ Reduction of unstable controllers not treated here