

Robust Control 2018

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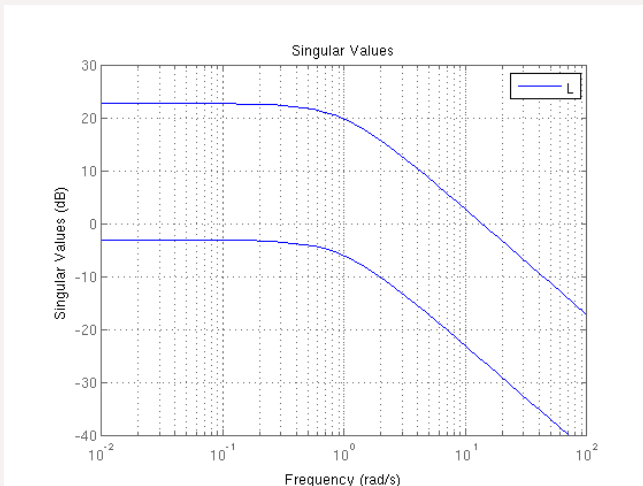
Plan of attack:

Today's topic: Evaluate \mathcal{H}_∞ based robust stability and performance claims.

- \mathcal{H}_∞ -norm performance specifications.
- The small gain theorem.
- Robust stability specifications.
- Proof of the small gain theorem:
 - Argument principle.
 - Loop transforms.
 - Instability theorems.

\mathcal{H}_∞ performance specifications

Suppose $L(j\omega)$ is given by the following:

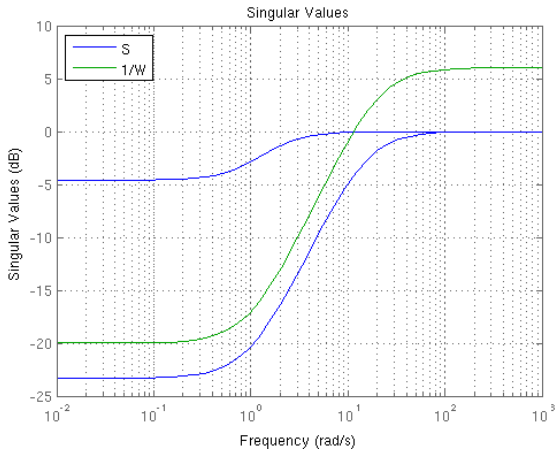


\mathcal{H}_∞ performance specifications

$$\text{Is } \left\| \frac{10(s/20+1)}{s+1} S(s) \right\|_\infty \leq 1??$$

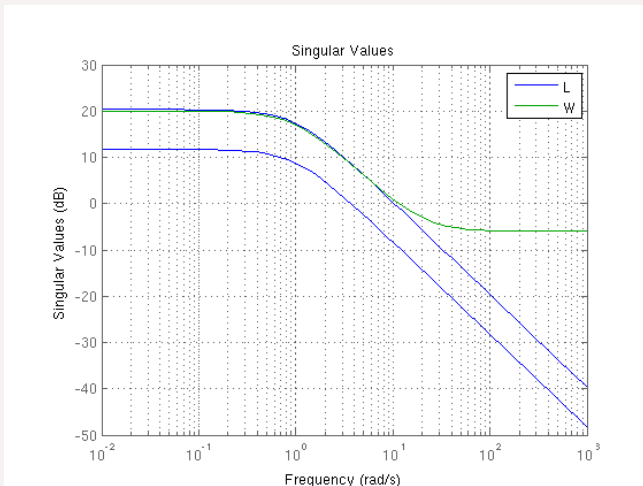
\mathcal{H}_∞ performance specifications

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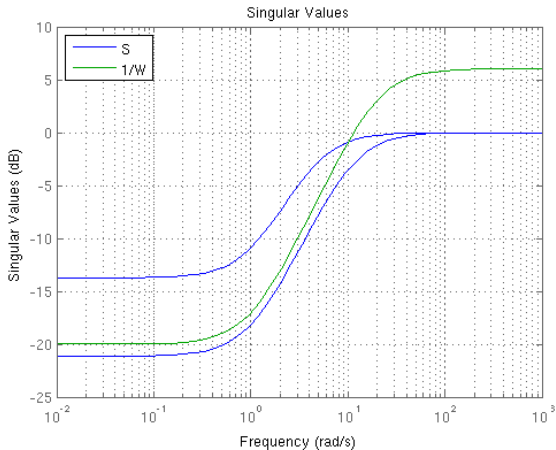


\mathcal{H}_∞ performance specifications

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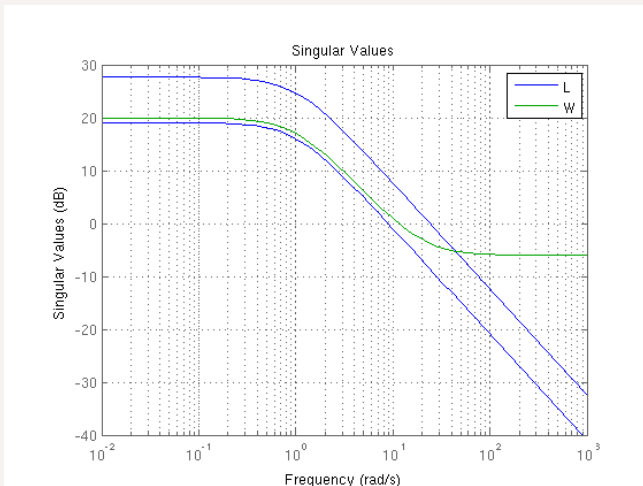


\mathcal{H}_∞ performance specifications

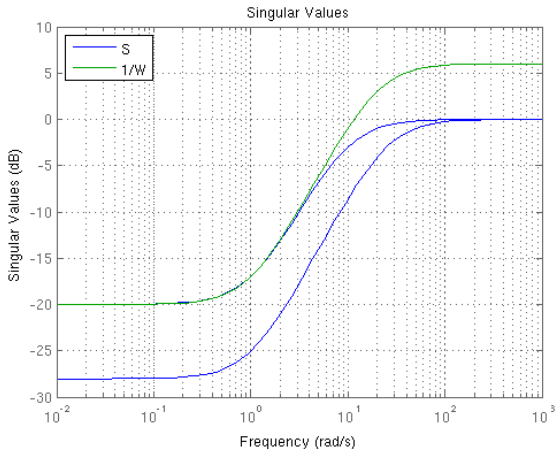


\mathcal{H}_∞ performance specifications

Suppose $L(j\omega)$ is given by the following:



\mathcal{H}_∞ performance specifications



The Bounded real Lemma

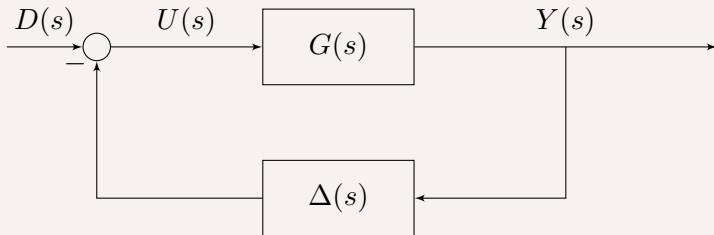
To check bounded gain, can use the singular value plot or the bounded real lemma.

Bounded real Lemma: Given any $G \in \mathcal{R}^{n \times m}$, the following are equivalent:

- (i) $\|G\|_\infty \leq 1$.
- (ii) For any minimal realisation of G there exists a $P \succ 0$ such that

$$\begin{bmatrix} A^T P + P A + C^T C & P B + C^T D \\ B^T P + D C^T & D^T D - I \end{bmatrix} \preceq 0.$$

The Small Gain Theorem



Given $G \in \mathcal{R}^{n \times m}$, the following are equivalent:

- (i) The feedback interconnection of G and Δ is stable for all $\Delta \in \mathcal{R}^{m \times n}$ such that $\|\Delta\|_\infty < 1$.
- (ii) $\|G\|_\infty \leq 1$.

The Small Gain Theorem

Can be generalized significantly. For example:

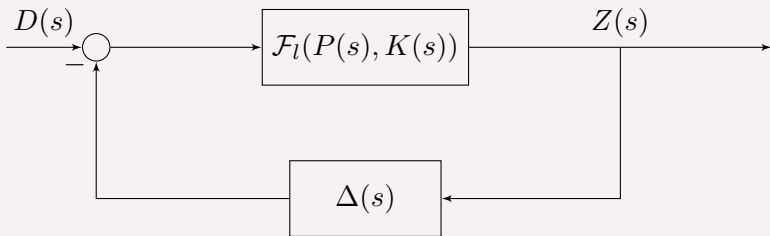
- Real rationality of G and Δ can be removed.
- Linearity of Δ can be removed.

Robust Stability Specifications

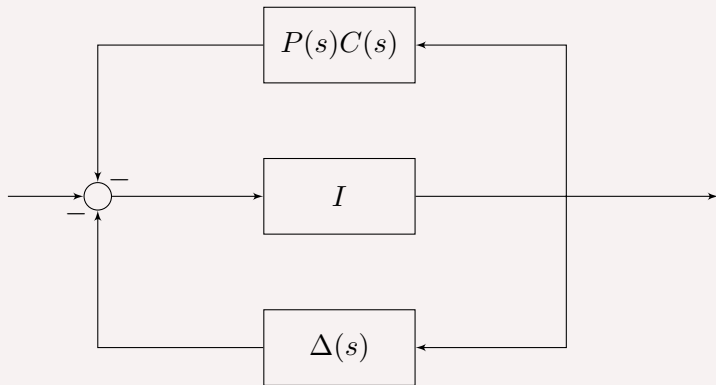
\mathcal{H}_∞ -norm performance specifications are robust stability specifications.

Robust Stability Specifications

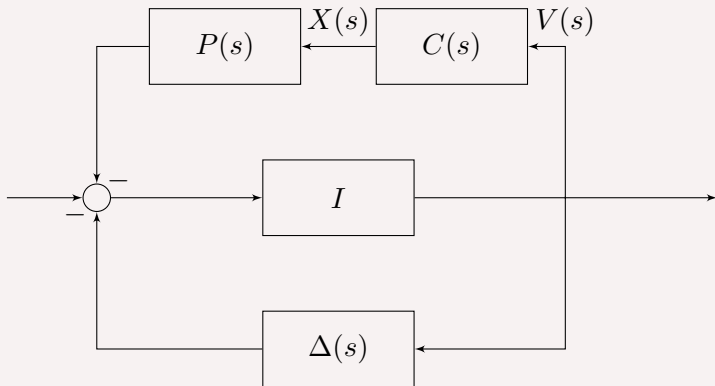
By the small gain theorem, $\|\mathcal{F}_l(P(s), K(s))\|_\infty \leq 1$ is equivalent to stability of the following for $\|\Delta\|_\infty < 1$:



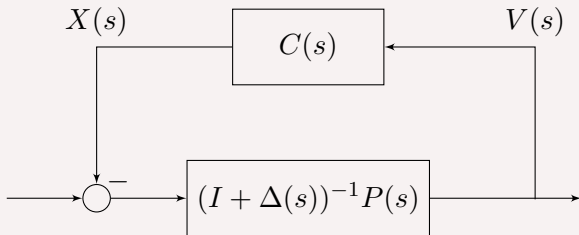
Robust Stability Specifications



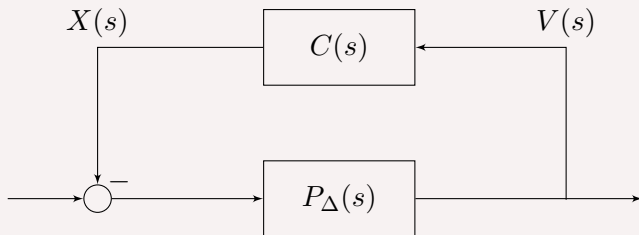
Robust Stability Specifications



Robust Stability Specifications



Robust Stability Specifications



$\|W(s)S(s)\|_{\infty} \leq 1 \implies$ stability for all

$$P_{\Delta}(s) \in \{Q(s) : Q(s) = (I + \Delta(s))^{-1}P(s), \|\Delta(s)\|_{\infty} < 1\}.$$

Robust Stability Specifications

Questions:

- 1 What are the robust stability equivalents of an \mathcal{H}_∞ -norm specification on each element of the gang of four?
- 2 What about the gang of four stability margin

$$b_{P,C} = \left\| \begin{bmatrix} P \\ I \end{bmatrix} (I - CP)^{-1} \begin{bmatrix} -C & I \end{bmatrix} \right\|_{\infty}^{-1} ?$$

Robust Stability Specifications

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Exercise

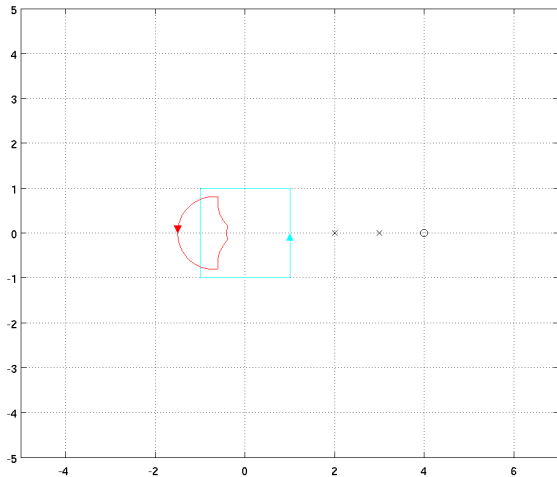
The Argument Principle

Consider

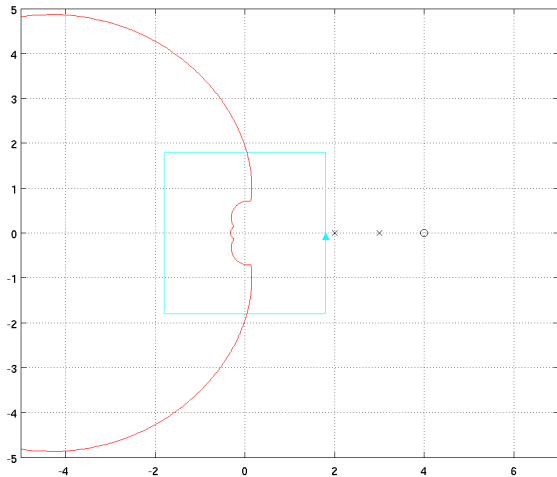
$$g(s) = \frac{s - 4}{(s - 2)(s - 3)}.$$

Let us evaluate $g(s)$ on a square contour.

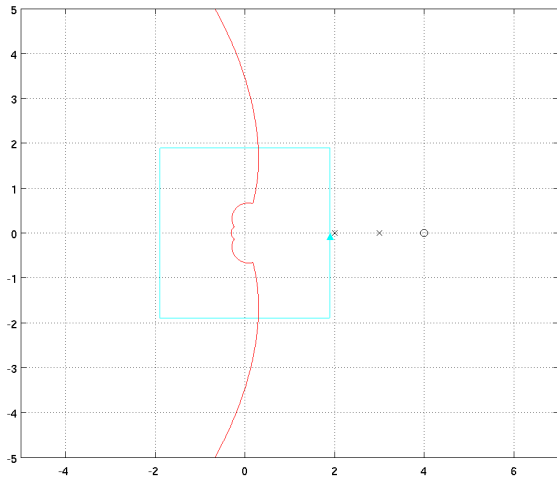
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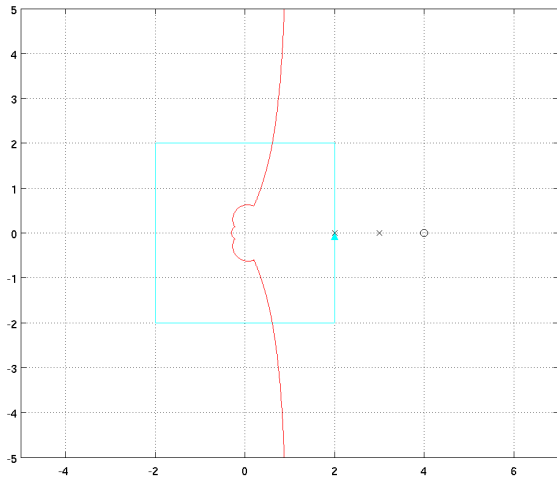
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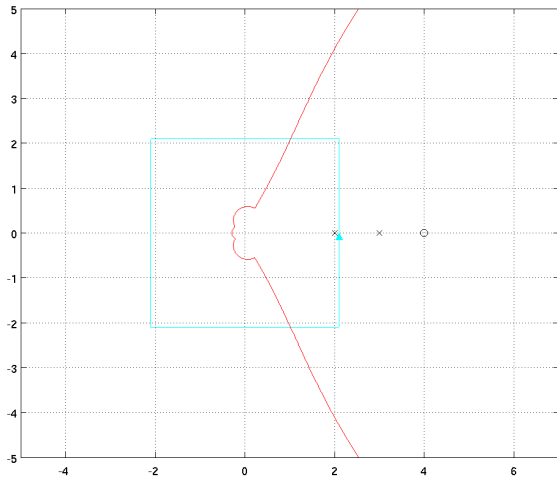
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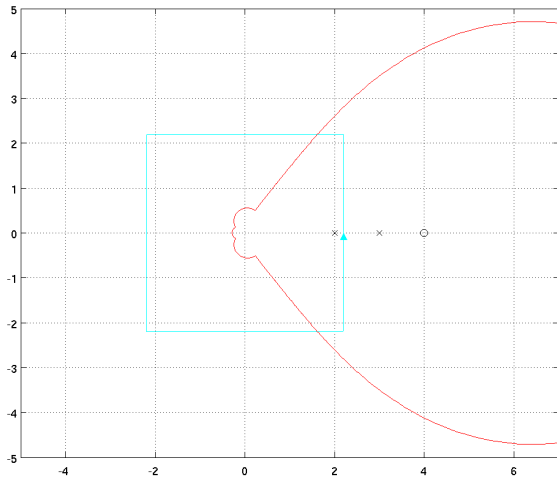
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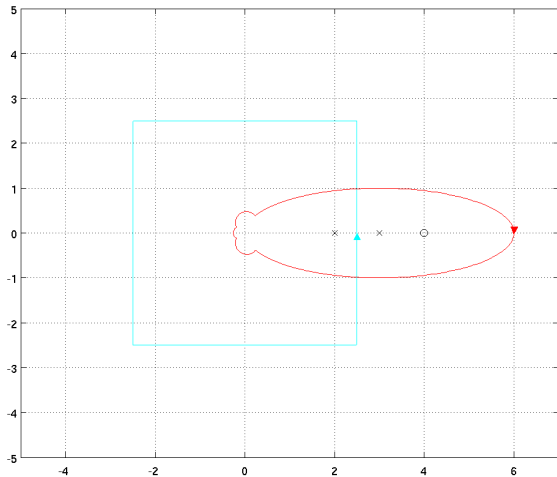
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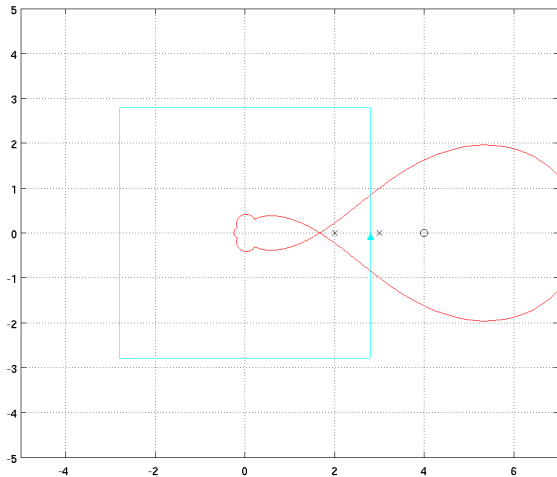
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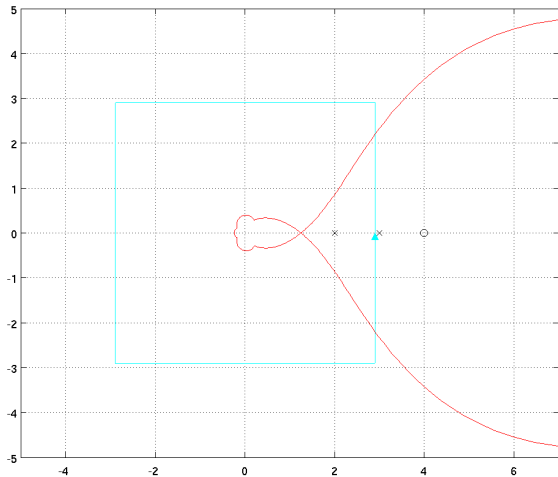
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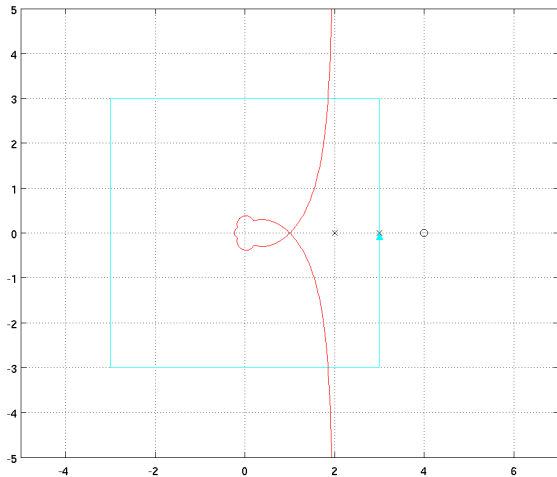
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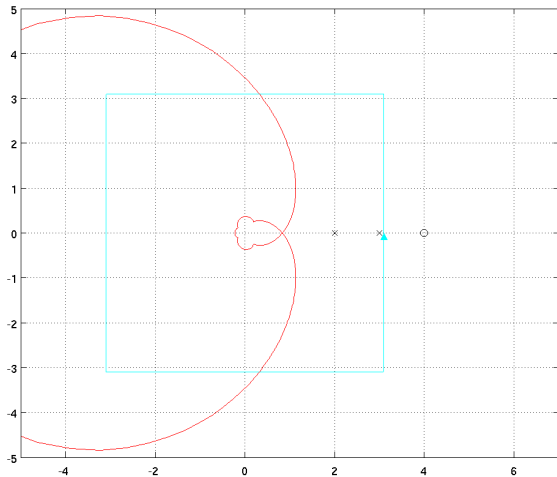
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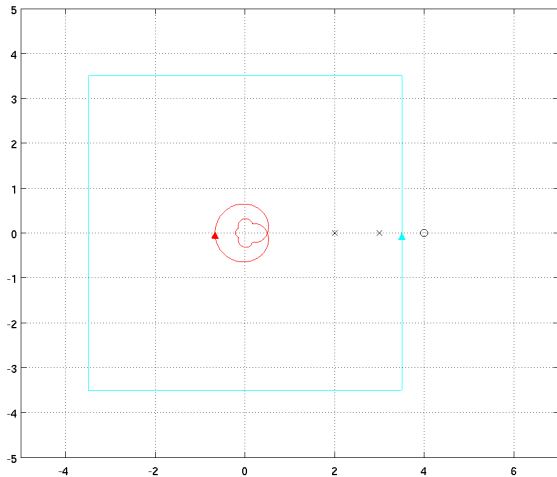
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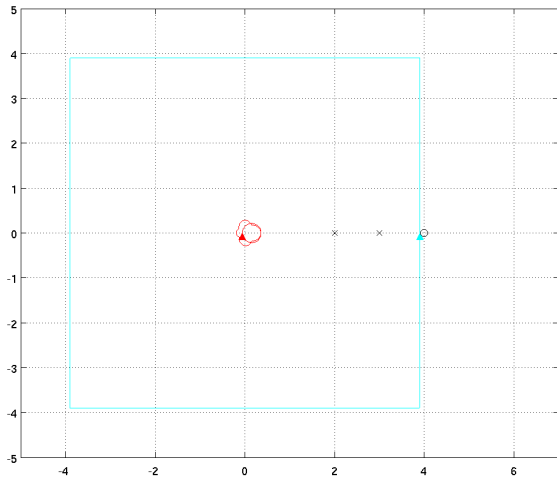
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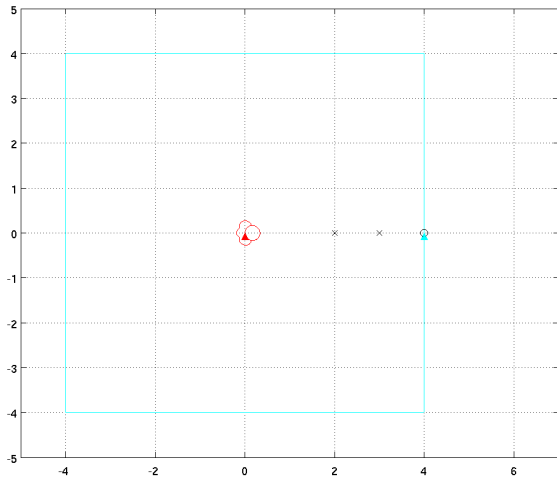
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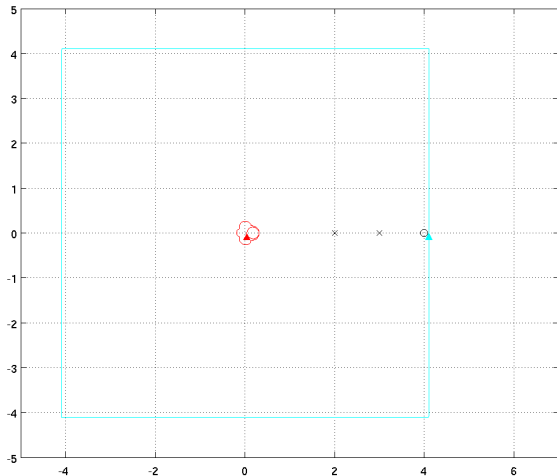
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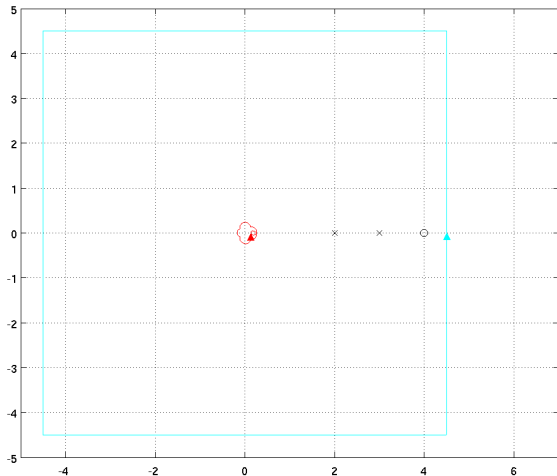
The Argument Principle



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The Argument Principle



The Argument Principle

Given $g(s) \in \mathcal{R}$ and a closed contour C ,

$$\text{w.n.o. } g(s) = Z - P$$

where Z, P are the number of zeros and poles of $g(s)$ contained in C .

The Argument Principle

Given $G(s) \in \mathcal{R}^{n \times n}$ and a closed contour C ,

$$\text{w.n.o. } \det(G(s)) = Z - P$$

where Z, P are the number of zeros and poles of $G(s)$ contained in C .

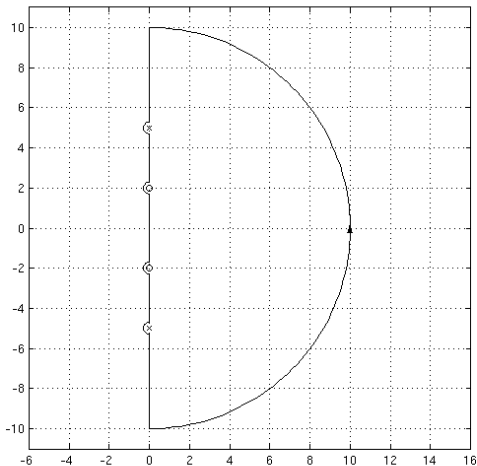
Proof of the Small Gain Theorem (ii) \Rightarrow (i)

We will use the argument principle to show that (ii) implies that

$$\det(I + G(s)\Delta(s)) \neq 0$$

for all s in the closed right half plane.

Proof of the Small Gain Theorem (ii) \Rightarrow (i)



The Nyquist contour.

Proof of the Small Gain Theorem (ii) \Rightarrow (i)

Consider

$$g_\lambda(s) = \det (I + \lambda G(s)\Delta(s)) .$$

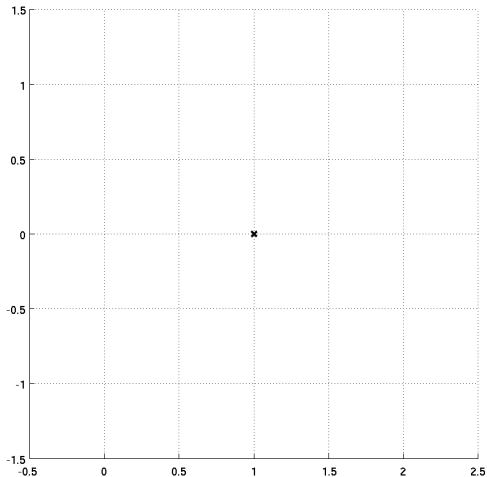
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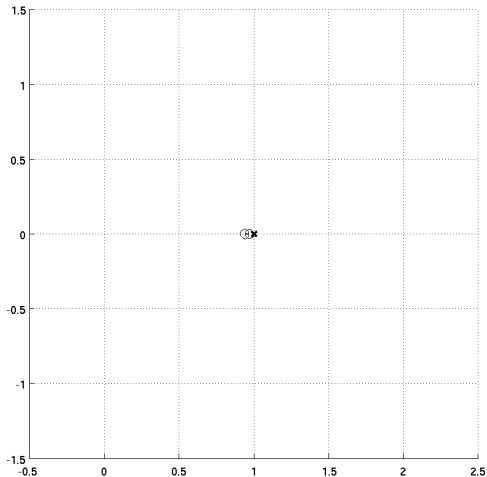
$$g_\lambda(s) = \det(I + \lambda G(s)\Delta(s)).$$

Now vary λ from 0 to 1.

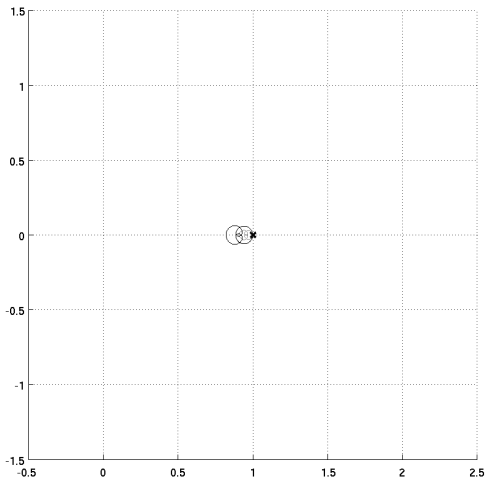
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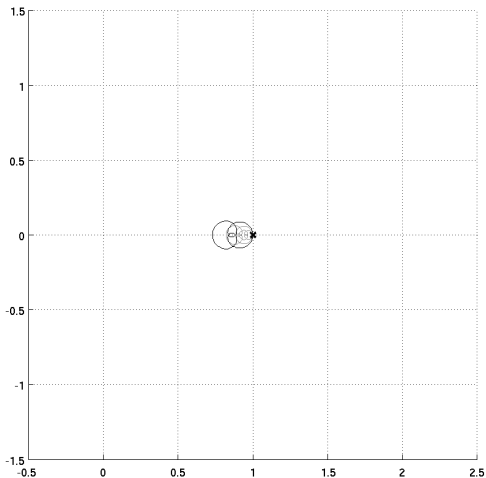
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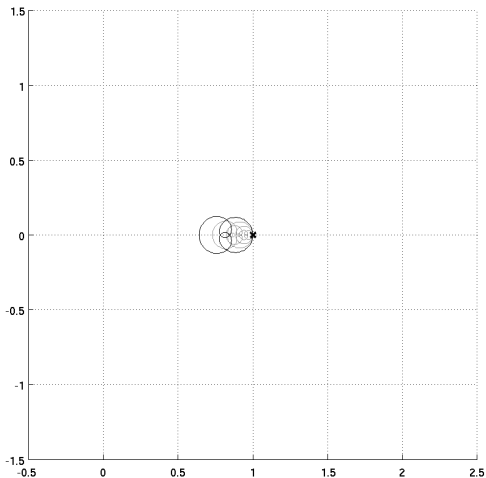
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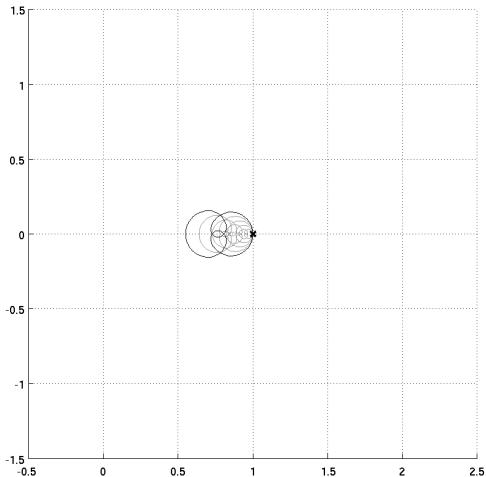
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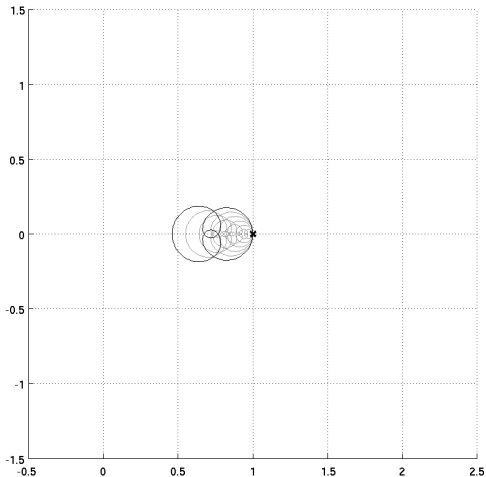
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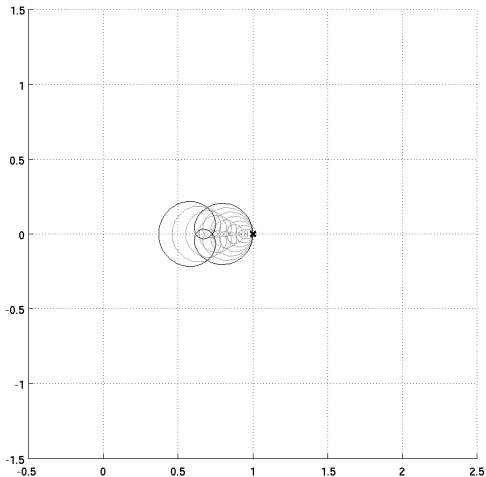
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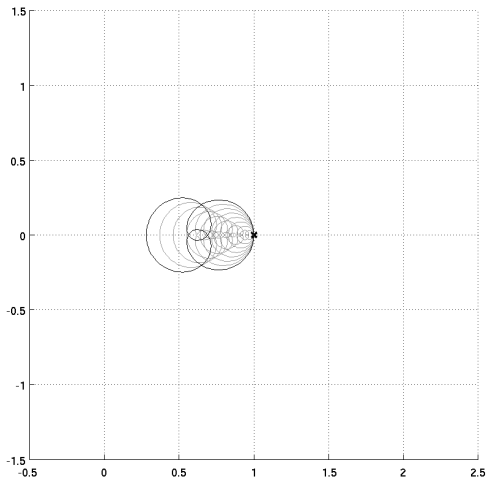
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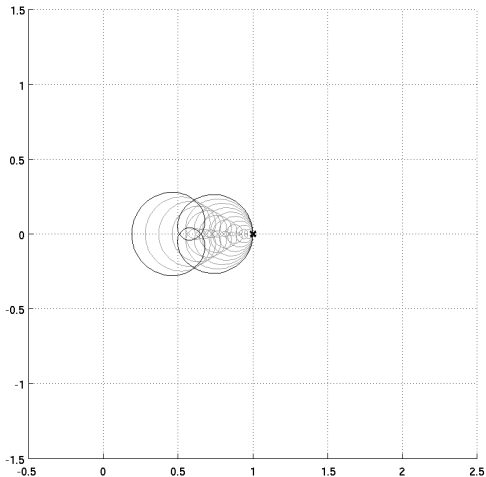
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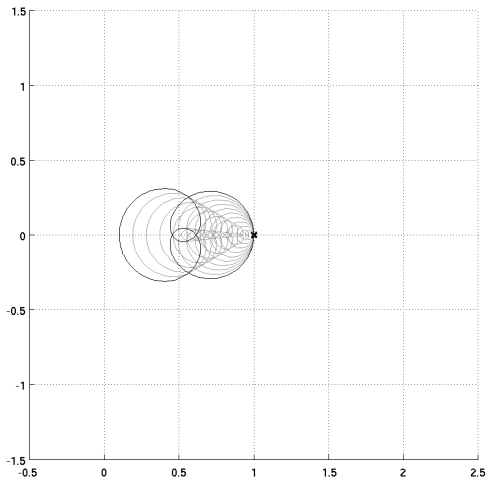
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Proof of the Small Gain Theorem (ii) \Rightarrow (i)



Proof of the Small Gain Theorem (ii) \Rightarrow (i)



Proof of the Small Gain Theorem (ii) \Rightarrow (i)

Recall that for $C \in \mathbb{C}^{n \times n}$

$$|\det C| = \prod_i \sigma_i(C).$$

Therefore if $\underline{\sigma}(C) > 0$, $\det C \neq 0$.

Proof of the Small Gain Theorem (ii) \Rightarrow (i)

Now recall that $\underline{\sigma}(A + B) \geq \underline{\sigma}(A) - \bar{\sigma}(B)$. Therefore since $\|G\|_\infty \leq 1$ and $\|\Delta\|_\infty < 1$,

$$|g_\lambda(j\omega)| \geq 1 - \lambda \bar{\sigma}(G(j\omega)\Delta(j\omega)) > 0.$$

Proof of the Small Gain Theorem (i) \Rightarrow (ii)

Therefore for every point on the contour, $g_\lambda(s) \neq 0$,

$$\Rightarrow \text{w.n.o. } g_0(s) = \text{w.n.o. } g_1(s).$$

Proof of the Small Gain Theorem (i) \Rightarrow (ii)

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By the argument principle, this implies that

$$0 = Z - P.$$

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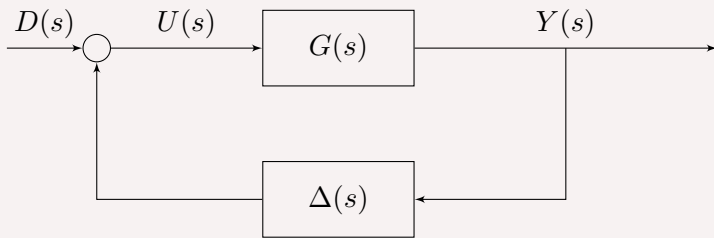
Since $P = 0$, \Rightarrow the interconnection is stable.

Proof of the Small Gain Theorem (i) \Rightarrow (ii)

Exercise

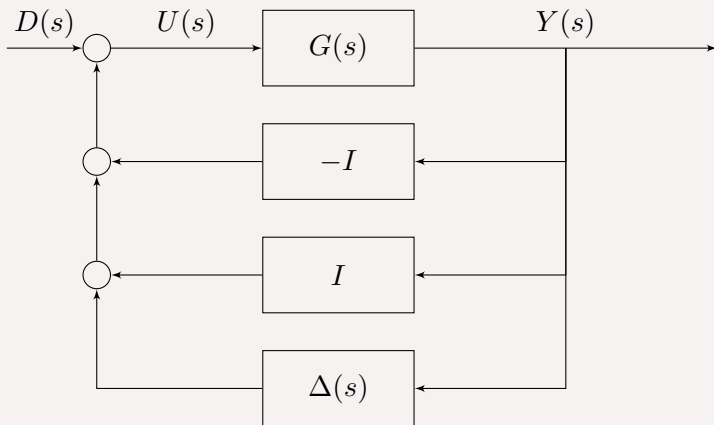
Loop Transforms

Can extend the small gain theorem using loop transforms:



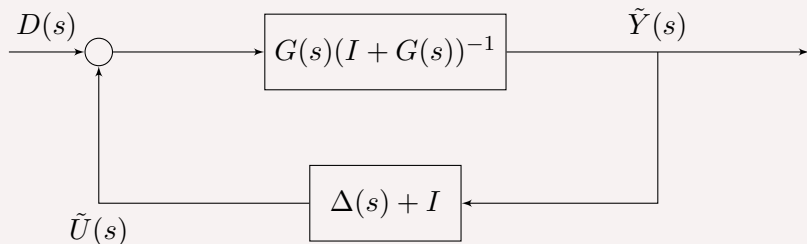
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Loop Transforms

Can extend the small gain theorem using loop transforms:



Stable if $\|G(s)(I + G(s))^{-1}\|_{\infty} \leq 1$ and $\|I + \Delta(s)\|_{\infty} < 1$.

Loop Transforms

- 1 Are we sure $(\tilde{Y}(s), \tilde{U}(s))$ bounded implies that $(Y(s), U(s))$ is bounded?
- 2 How can we interpret such loop transforms?

Using the Projective Line

Recall from last time that we used the projective line to understand $G(s)(I + G(s))^{-1}$. Use the chain description:

$$\begin{bmatrix} I & 0 \\ I & I \end{bmatrix} \begin{bmatrix} G(s) \\ I \end{bmatrix}$$

Using the Projective Line

Signal interpretation:

$$\begin{bmatrix} \tilde{Y} \\ \tilde{U} \end{bmatrix} = \begin{bmatrix} I & 0 \\ I & I \end{bmatrix} \begin{bmatrix} Y \\ U \end{bmatrix}, \quad \begin{bmatrix} Y \\ U \end{bmatrix} = \begin{bmatrix} G \\ I \end{bmatrix} U.$$

- 1 The first equation describes the loop transform.
- 2 In the second, U generates the 'behaviour' (use coprime factorisation for unstable G)

Using the Projective Line

Let $M \in \mathcal{R}^{n \times n}$ define a loop transform

$$\begin{bmatrix} \tilde{Y}(s) \\ \tilde{U}(s) \end{bmatrix} = M(s) \begin{bmatrix} Y(s) \\ U(s) \end{bmatrix}.$$

Boundedness of $(\tilde{Y}(s), \tilde{U}(s))$ is equivalent to boundedness of $(Y(s), U(s))$ if and only if $M, M^{-1} \in \mathcal{RH}_\infty$.

The Passivity Theorem

Small gain of $Y(s) = G(s)U(s)$ is equivalent to

$$\begin{bmatrix} Y(s) \\ U(s) \end{bmatrix}^* \begin{bmatrix} -I & 0 \\ 0 & I \end{bmatrix} \begin{bmatrix} Y(s) \\ U(s) \end{bmatrix} \geq 0.$$

The Passivity Theorem

Suppose

$$\begin{bmatrix} Y(s) \\ U(s) \end{bmatrix} = \frac{1}{\sqrt{2}} \begin{bmatrix} I & I \\ -I & I \end{bmatrix} \begin{bmatrix} \tilde{Y}(s) \\ \tilde{U}(s) \end{bmatrix}?$$

The Passivity Theorem

Small gain is equivalent to:

$$\left(\frac{1}{\sqrt{2}} \begin{bmatrix} I & I \\ -I & I \end{bmatrix} \begin{bmatrix} \tilde{Y}(s) \\ \tilde{U}(s) \end{bmatrix} \right)^* \begin{bmatrix} -I & 0 \\ 0 & I \end{bmatrix} \frac{1}{\sqrt{2}} \begin{bmatrix} I & I \\ -I & I \end{bmatrix} \begin{bmatrix} \tilde{Y}(s) \\ \tilde{U}(s) \end{bmatrix} \geq 0.$$

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The Passivity Theorem

Small gain is equivalent to:

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$$\implies \int_{-\infty}^{\infty} \tilde{y}^T(t) \tilde{u}(t) dt \geq 0.$$

The Passivity Theorem

What about the transfer function given by the chain description:

$$\frac{1}{\sqrt{2}} \begin{bmatrix} I & I \\ -I & I \end{bmatrix}^{-1} \begin{bmatrix} G(s) \\ I \end{bmatrix}?$$

The Passivity Theorem

What about the transfer function given by the chain description:

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Exercise

The Passivity Theorem

- Small gain of the pair (y, u) is equivalent to passivity of (\tilde{y}, \tilde{u}) .
- (y, u) are called scattering variables in this context.
- Connections to IQCs, the KYP lemma, chain scattering, ...

Instability Theorems

Observe that in our proof of the small gain theorem we only used our \mathcal{H}_∞ -norm bound to bound the largest singular value of G on the Nyquist contour.

Instability Theorems

\mathcal{L}_∞ is the space of essentially bounded measurable functions on the imaginary axis, with norm

$$\|f(j\omega)\|_{\mathcal{L}_\infty} = \operatorname{ess\,sup}_{\omega \in \mathbb{R}} |f(j\omega)|.$$

Instability Theorems

For real rational functions:

$$\|f(j\omega)\|_{\mathcal{L}_\infty} = \sup_{\omega \in \mathbb{R}} |f(j\omega)|.$$

Instability Theorems

For matrices of real rational functions:

$$\|F(j\omega)\|_{\mathcal{L}_\infty} = \sup_{\omega \in \mathbb{R}} \bar{\sigma}(F(j\omega)).$$

Instability Theorems

By our previous argument, $\|G(j\omega)\|_{\mathcal{L}_\infty} \leq 1$ and $\|\Delta(j\omega)\|_{\mathcal{L}_\infty} < 1$ is sufficient to conclude that

$$\text{w.n.o. } \det(I + G\Delta) = 0 = Z - P.$$

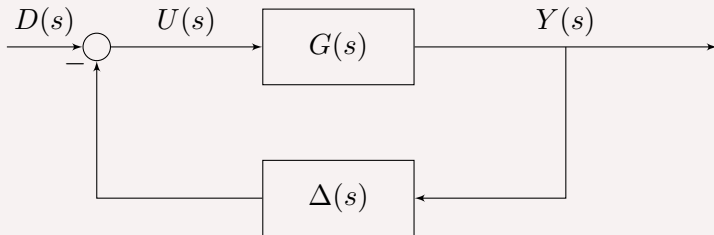
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$$\text{w.n.o. } \det(I + G\Delta) = 0 = Z - P.$$

\implies that the number of closed loop poles equals the number of open loop poles.

The Small Gain Theorem (2)



Given $G \in \mathcal{R}^{n \times m}$, the following are equivalent:

- (i) The feedback interconnection of G and Δ has P unstable poles for all $\Delta \in \mathcal{R}^{m \times n}$ such that $\|\Delta\|_\infty < 1$.
- (ii) $\|G\|_{\mathcal{L}_\infty} \leq 1$ and G has P unstable poles.

The Small Gain Theorem (2)

- 1 Can broaden applicability with the same projective line arguments.
- 2 Real rational and linear requirements can be relaxed.
- 3 Can check $\|G\|_{\mathcal{L}_\infty} \leq 1$ using LMIs.