Stability of Sampled-Data Control Loops under Input and Output Jitter

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The Bigger Picture

Ultimate goal: Optimal control and scheduling co-synthesis of networked embedded control systems



- Guarantee stability of all control loops
- Minimize performance degradation due to scheduling

Textbook Sampled-Data Control



[Åström & Wittenmark, 1997]

Delay and Jitter due to Scheduling

One control task:



- Constant input–output delay
- No jitter
- Textbook analysis applicable

Delay and Jitter due to Scheduling

Many control tasks (EDF scheduling):



- Both input and output jitter
- Bounds on the jitter can be found by schedulability analysis
- What about stability and performance?

Jitter Definitions



Input jitter: $J_i = \max_k \tau_i^k - \min_k \tau_i^k$ Output jitter: $J_o = \max_k \tau_o^k - \min_k \tau_o^k$

The constant delay part can be included in the plant

Analysis of Output Jitter

C.-Y. Kao and B. Lincoln: "Simple Stability Criteria for Systems with Time-Varying Delays," *Automatica* **40**:8, 2004.

- Periodic sampling (no input jitter)
- Linear plant P(s) and controller K(z)
- Effect of output jitter modeled as a plant input disturbance:



Analysis of Output Jitter

Assume $J_o \leq h$. Loop transformation:



 Λ_h : time-varying gate function with gain

$$||\Lambda_h||_{L_2} = \sqrt{J_o/h}$$

Analysis of Output Jitter

Theorem [Kao and Lincoln]:

Closed-loop system stable for any $\{ au_o^k\}\in [0,J_o]$ if

$$\left|\frac{P_{\text{alias}}(\omega)K(e^{i\omega})}{1+P_{\text{ZOH}}(e^{i\omega})K(e^{i\omega})}\right| < \frac{1}{\sqrt{J_o/h}\left|e^{i\omega}-1\right|} \quad \forall \omega \in [0,\pi]$$

where

$$P_{
m alias}(\omega) = \sqrt{\sum_{k=-\infty}^{\infty} \left| P\left(i(\omega+2\pi k)rac{1}{h}
ight)
ight|^2}$$

Output Jitter – Example

Plant (inverted pendulum):

$$P(s) = \frac{1}{(1+s10)(1-s10)}$$

Controller with sampling period h = 200 ms:

$$K(z) = \frac{3.36z(z - 0.969)}{z^2 - 1.77z + 0.794}$$

Fixed delay margin: 200 ms

Jitter margin by Kao and Lincoln: 153 ms

How conservative is it?

- Assume that the delay can only take on N discrete values between 0 and J_o
- Each delay value gives rise to a closed-loop matrix Φ_i
- *n*-step stability can be asserted by finding a quadratic Lyapunov function for the set of all closed-loop matrices

$$\{\Phi_1,\ldots,\Phi_N\}^n$$

N = 30, n = 2 gives approximate jitter margin of 189 ms

An Improved Analysis

Modified loop transformation:



- Block the transmission of signals with delay outside [0, *J_o*] using a fixed gate function Λ(*J_o*)
- Calculate the gain by fast sampling/fast hold approximation

New jitter margin: 176 ms

Accounting for Input Jitter

Assume that P(s) has the state-space realization

$$\dot{x}(t) = Ax(t) + Bu(t)$$
$$y(t) = Cx(t)$$

An input delay $0 \le \tau_i^k \le J_i$ creates an effective error in the measurement according to

$$\begin{split} e_k &= Cx(kh + \tau_i^k) - Cx(kh) \\ &= C\Big(\Phi(\tau_i^k)x_k + \Gamma(\tau_i^k)u_{k-1}\Big) - Cx_k \\ &= \underbrace{\Big(C(\Phi(\tau_i^k) - I) \quad C\Gamma(\tau_i^k)\Big)}_{\Delta_S} \begin{pmatrix} x_k \\ u_{k-1} \end{pmatrix} \end{split}$$

 Δ_S is a time-varying, memory-less MISO operator

Accounting for Sampling Jitter

Loop transformation with two uncertainty blocks:



Small gain theorem

Example Revisited



Conclusions

- Multitasking networked control systems give rise to input jitter and output jitter
- Fast stability and performance analysis is needed to carry out efficient co-synthesis of large networked embedded control systems
- Simple extensions of previous results by Kao and Lincoln
 - Reduced conservativeness for sub-sample jitter
 - Input jitter treated as another uncertainty